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# (12) United States Patent

#### Ebisuno et al.

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#### (54) **DISPLAY DEVICE**

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U.S.C. 154(b) by 337 days.

This patent is subject to a terminal dis-

claimer.

(21) Appl. No.: 13/768,217

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(51) **Int. Cl.** 

G09G 3/30 (2006.01) H05B 33/08 (2006.01) G09G 3/32 (2006.01)

(52) U.S. Cl.

(58) Field of Classification Search

See application file for complete search history.

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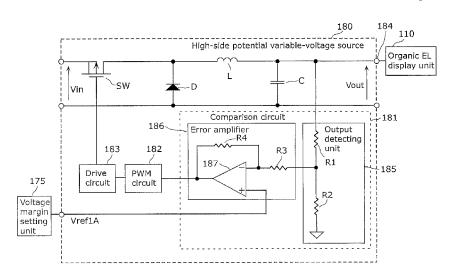
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#### (57) ABSTRACT

A display device includes: a high-side potential variablevoltage source which outputs a high-side output potential and a low-side potential variable-voltage source which outputs a low-side output potential; an organic EL display unit in which pixels are arranged; a high-side potential difference detecting circuit which detects a high-side potential applied to a first pixel and a low-side potential difference detecting circuit which detects a low-side potential applied to a second pixel; a high-side potential voltage margin setting unit and a lowside potential voltage setting unit which regulate the output potential of the high-side potential variable-voltage source and the low-side potential variable-voltage source to set a potential difference between the high-side potential of the first pixel and a reference potential to a predetermined potential difference and set a potential difference between the lowside potential of the second pixel and a reference potential to a predetermined potential difference; and a signal processing circuit.

## 13 Claims, 25 Drawing Sheets

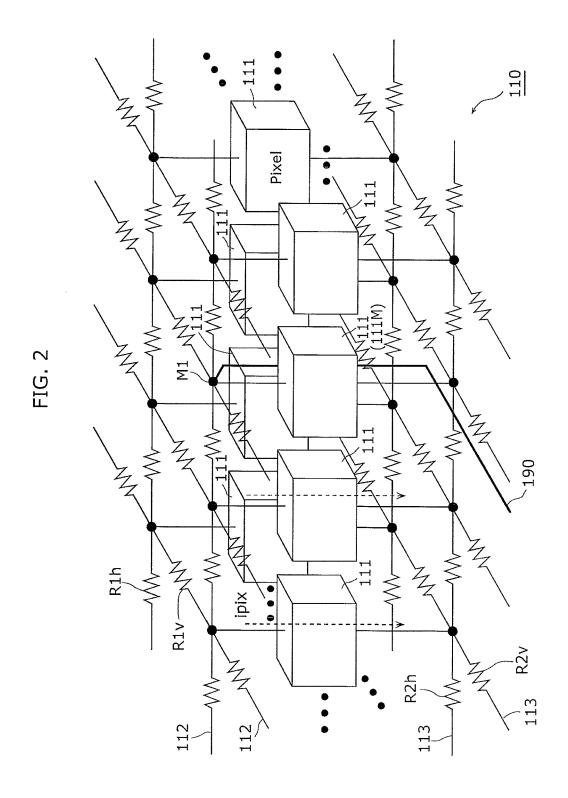


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High-side potential variable-voltage source Display device 180 112 difference detecting High-side potential \_175 190 Voltage margin setting unit circuit 120 7165 Organic EL display unit Signal processing circuit Data line driving circuit / **M1** Video data 140 130 Control scan driving circuit circuit

FIG. 1



VTFT VEL High-side potential difference detecting circuit 170 Monitor pixel -190 $\overline{\mathsf{M}}$ FIG. 3 123 112 124 Data line driving circuit 120 -122Write scan driving -circuit 130 voltage source 180 High-side potential variable-

FIG. 4

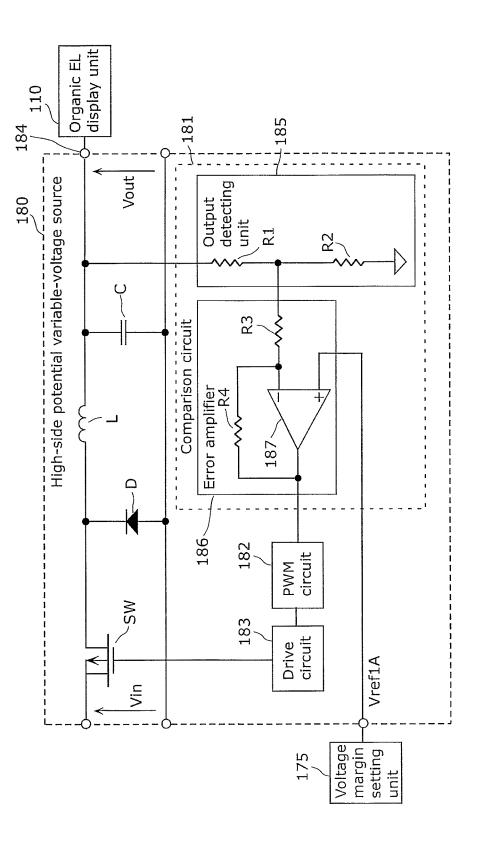


FIG. 5

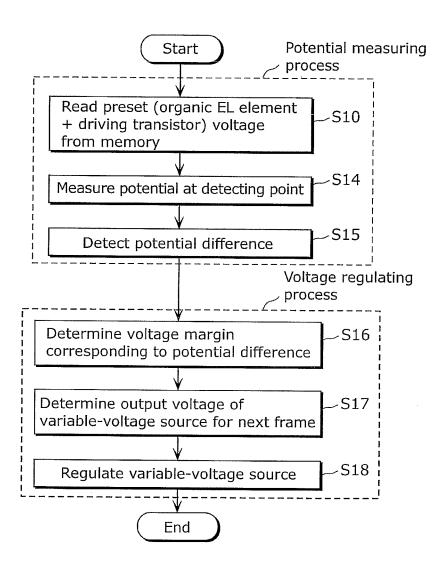


FIG. 6

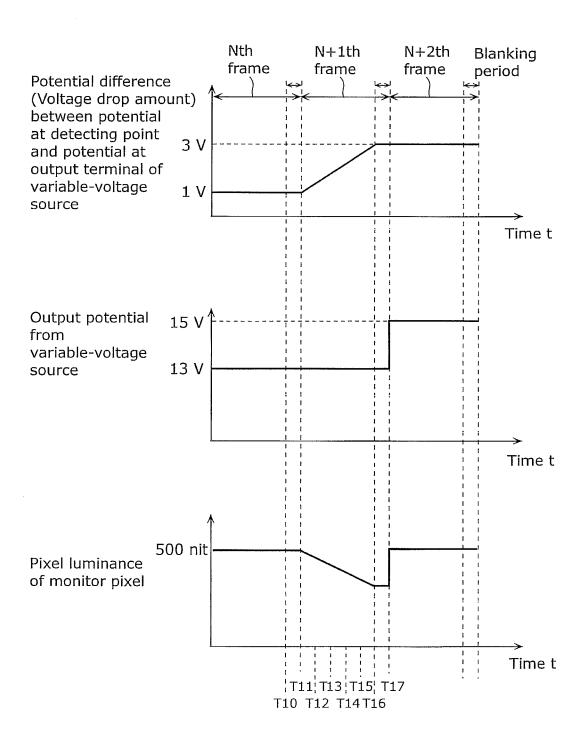
Video data (Gradation level)	Required voltage (Red)	Required voltage (Green)	Required voltage (Blue)
255	11.2	12.2	8.4

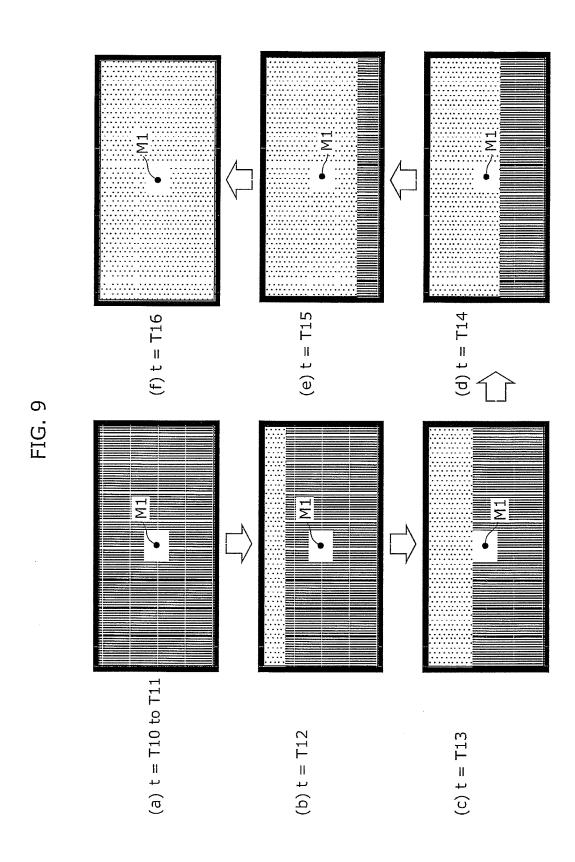
FIG. 7

Potential difference value [V]	Voltage drop margin	
0.0	0.0	
0.2	0.2	
0.4	0.4	
0.6	0.6	
:	•	
3.4	3.4	
3.6	3.6	
:	:	
5.6	5.6	
5.8	5.8	
6.0	6.0	

FIG. 8

Nov. 10, 2015





Display device High-side potential variable-voltage High-side potential difference detecting source 7190 circuit 160 150 Peak signal detecting circuit Data line driving circuit Signal processing circuit Organic EL display unit  $\frac{1}{2}$ Video data 140 130 Control scan driving circuit circuit Write

FIG. 1(

Organic EL display unit 110 181 -185 184 Output detecting unit Vout High-side potential variable-voltage source 180 R R2R3 Error amplifier 187 186 182 circuit PWM 183 Drive circuit Vref1 Κin Signal processing-circuit 160

FIG. 12

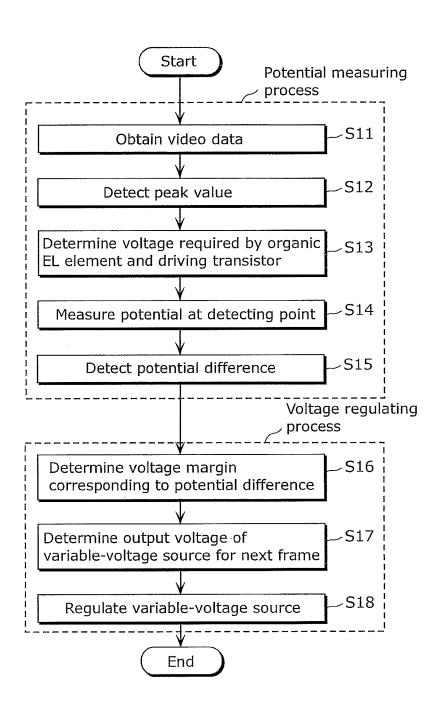
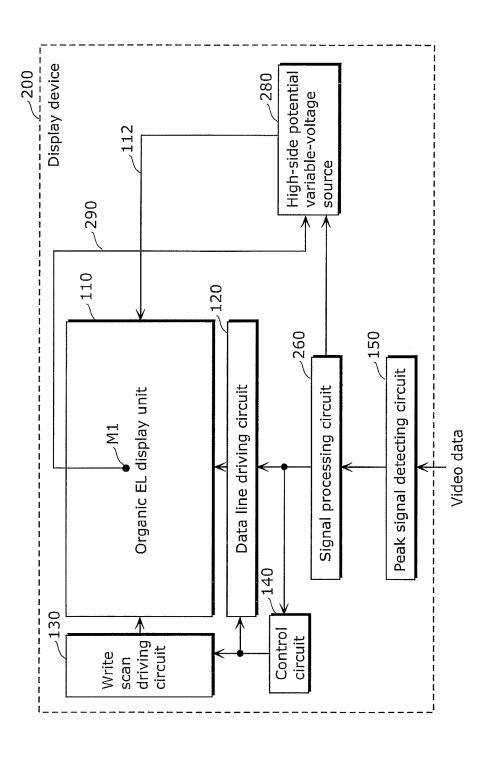


FIG. 13

Video data (Gradation level)	Required voltage (Red)	Required voltage (Green)	Required voltage (Blue)
0	4	4.2	3.5
1	4.1	4.3	3.5
2	4.1	4.4	3.6
3	4.2	4.5	3.6
	:	•	•
176	8.3	9.6	6.7
177	8.5	9.9	6.9
:	•	•	•
253	10.5	11.4	8.2
254	10.8	11.8	8.3
255	11.2	12.2	8.4

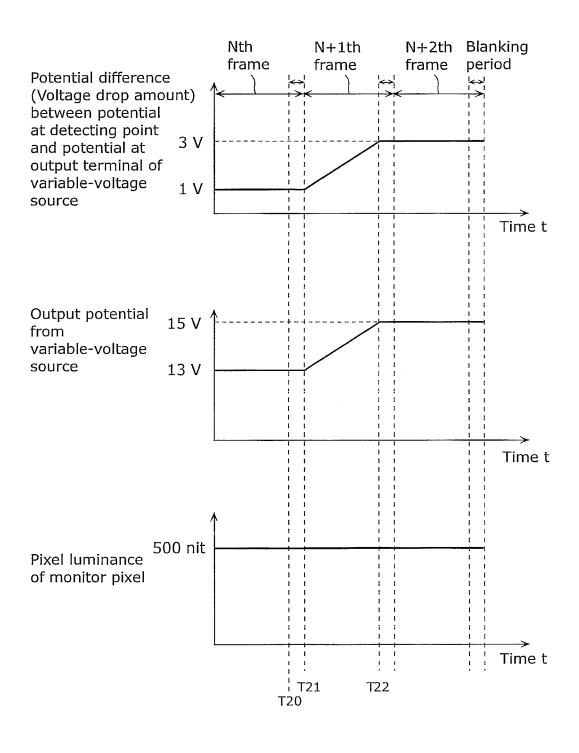
FIG. 14



Nov. 10, 2015

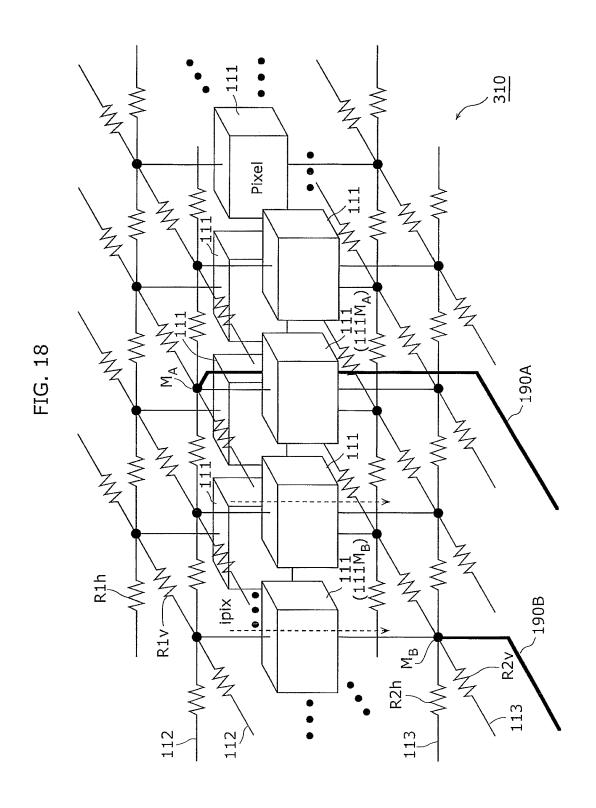
Organic EL display unit 110 290 112 281 185 184 Vout Output detecting unit High-side potential variable-voltage source 280 RI R2 R3 Comparison circuit Error amplifier FIG. 15 187 182 circuit PWM NS. 183 Drive circuit Vref2 Signal processing-circuit 260

FIG. 16



180B 300 variable-voltage Display device Low-side potential voltage margin Low-side potential 175B 170B setting unit source 190B potential difference detecting Low-side circuit 80A variable-voltage High-side potential High-side potential voltage margin 175A source 190A setting unit difference High-side detecting potential circuit 310 120 Organic EL display unit Signal processing circuit Data line driving circuit 165 Video data  $M_{\rm B}$ 130 Control scan driving circuit circuit 140 Write

FIG. 17



Nov. 10, 2015

FIG. 19A

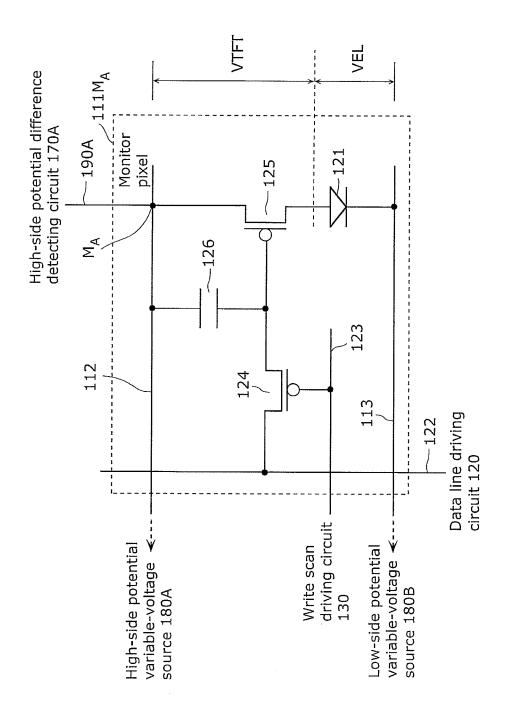


FIG. 19B

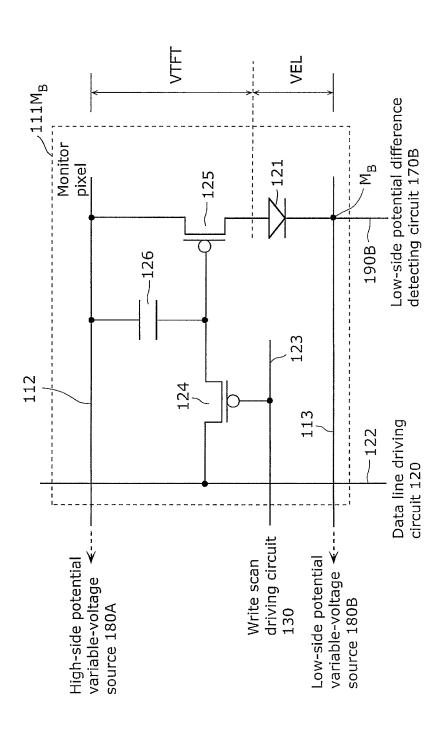
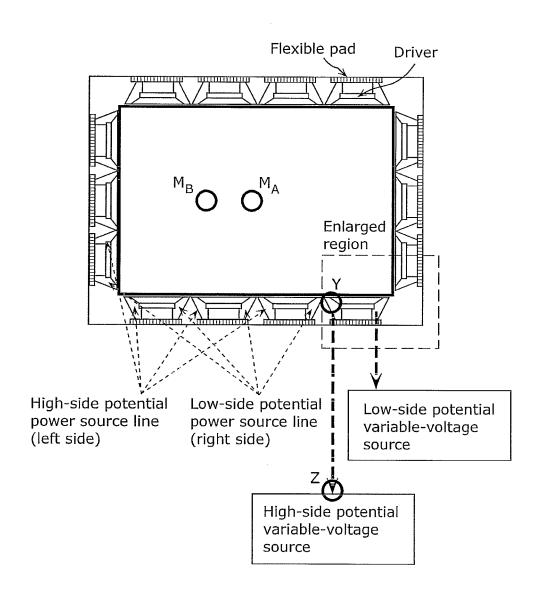


FIG. 20A



High-side potential Low-side potential variable-voltage variable-voltage Current path (+) Outside display region Current path (-) == source source component Resistance in current path Inside display region | Σ

TG. 201

FIG. 21

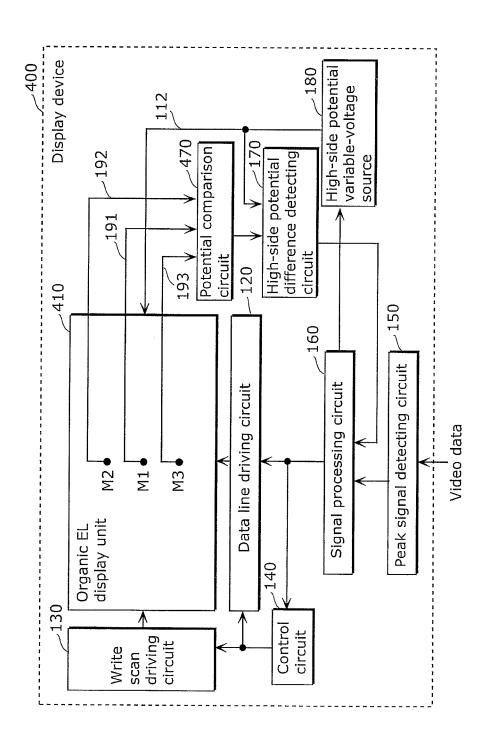


FIG. 22

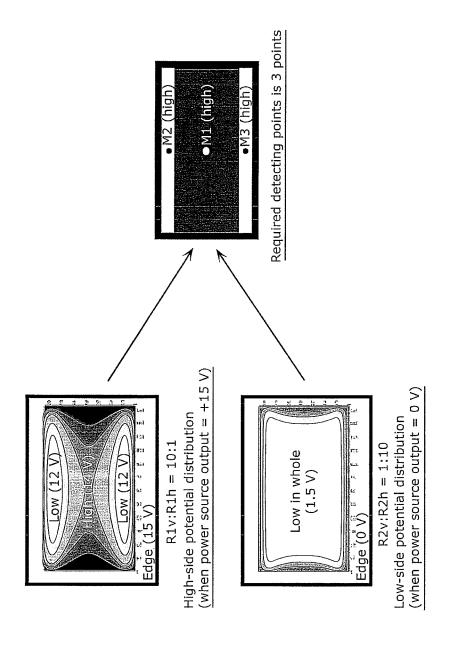


FIG. 23

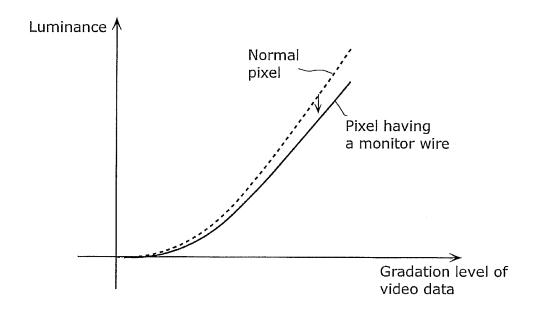


FIG. 24

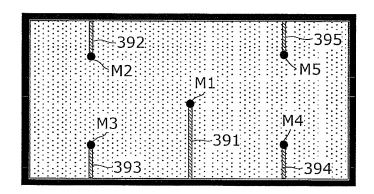


FIG. 25

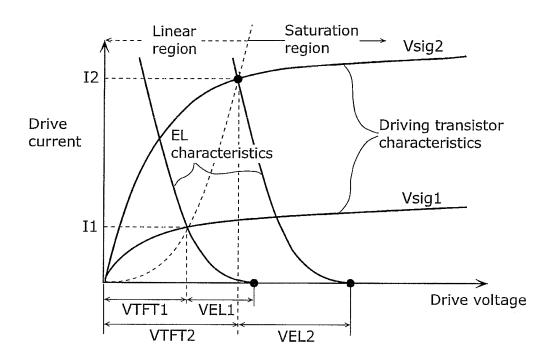
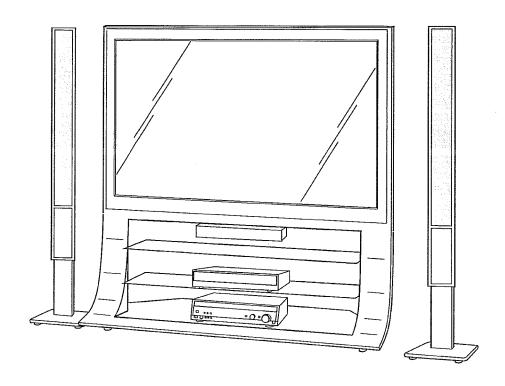


FIG. 26



### DISPLAY DEVICE

## CROSS REFERENCE TO RELATED APPLICATION

This is a continuation application of PCT Patent Application No. PCT/JP2011/003424 filed on Jun. 16, 2011, designating the United States of America. The entire disclosure of the above-identified application, including the specification, drawings and claims are incorporated herein by reference in its entirety.

#### **FIELD**

One or more exemplary embodiments disclosed herein relate generally to active-matrix display devices which use current-driven luminescence elements represented by organic electroluminescence (EL) elements, and more particularly to a display device having excellent power consumption reducing effect.

#### BACKGROUND

In general, the luminance of an organic electroluminescence (EL) element is dependent upon the drive current supplied to the element, and the luminance of the element increases in proportion to the drive current. Therefore, the power consumption of displays made up of organic EL elements is determined by the average of display luminance. Specifically, unlike liquid crystal displays, the power consumption of organic EL displays varies significantly depending on the displayed image.

For example, in an organic EL display, the highest power consumption is required when displaying an all-white image, whereas, in the case of a typical natural image, power consumption which is approximately 20 to 40 percent of that for all-white is considered to be sufficient.

However, because power source circuit design and battery capacity entail designing which assumes the case where the power consumption of a display becomes highest, it is necessary to consider power consumption that is 3 to 4 times that for the typical natural image, and thus becoming a hindrance to the lowering of power consumption and the miniaturization of devices.

Consequently, there is conventionally proposed a technique which suppresses power consumption with practically no drop in display luminance, by detecting the peak value of video data and regulating the cathode voltage of the organic EL elements based on such detected data so as to reduce power source voltage (for example, see Patent Literature 50 (PTL) 1).

#### CITATION LIST

## Patent Literature

[PTL 1] Japanese Unexamined Patent Application Publication No. 2006-065148

## SUMMARY

#### Technical Problem

Now, since an organic EL element is a current-driven element, current flows through a power source wire and a voltage 65 drop which is proportionate to the wire resistance occurs. As such, the power supply voltage to be supplied to the display is

2

set by adding a voltage drop margin for compensating for a voltage drop. In the same manner as the previously described power source circuit design and battery capacity, since the power drop margin for compensating for a voltage drop is set assuming the case where the power consumption of the display becomes highest, unnecessary power is consumed for typical natural images.

In a small-sized display intended for mobile device use, panel current is small and thus, compared to the voltage to be consumed by pixels, the power drop margin for compensating for a voltage drop is negligibly small. However, when current increases with the enlargement of panels, the voltage drop occurring in the power source wire no longer becomes negligible.

However, in the conventional technique in the above-mentioned Patent Reference 1, although power consumption in each of the pixels can be reduced, the power drop margin for compensating for a voltage drop cannot be reduced, and thus the power consumption reducing effect for household large-sized display devices of 30-inches and above is insufficient.

One non-limiting and exemplary embodiment provides a display device having excellent power consumption reducing effect.

#### Solution to Problem

In one general aspect, the techniques disclosed herein feature a display device including: a power supplying unit configured to output at least one of a high-side output potential and a low-side output potential; a display unit in which pixels are arranged and which receives power supply from the power supplying unit; a voltage detecting unit configured to detect at least one of a high-side applied potential to be applied to at least one of the pixels in the display unit and a low-side applied potential to be applied to at least one of the pixels in the display unit which is the same as or different from the at least one of the pixels to which the high-side applied potential is to be applied; and a voltage regulating unit configured to regulate at least one of the high-side output potential and the low-side output potential outputted by the power supplying unit to set, to a predetermined potential difference, a potential difference between (i) at least one of the high-side applied potential and the low-side applied potential and (ii) a refer-

These general and specific aspects may be implemented using a system, a method, an integrated circuit, a computer program, or a computer-readable recording medium such as a CD-ROM, or any combination of systems, methods, integrated circuits, computer programs, or computer-readable recording media.

Additional benefits and advantages of the disclosed 50 embodiments will be apparent from the Specification and Drawings. The benefits and/or advantages may be individually obtained by the various embodiments and features of the Specification and Drawings, which need not all be provided in order to obtain one or more of such benefits and/or advantages.

#### Advantageous Effects

The display device according to one or more exemplary 60 embodiments or features disclosed herein enables the implementation of a display device having excellent power consumption reducing effect.

#### BRIEF DESCRIPTION OF DRAWINGS

These and other advantages and features will become apparent from the following description thereof taken in con-

junction with the accompanying Drawings, by way of nonlimiting examples of embodiments disclosed herein.

FIG. 1 is a block diagram showing an outline configuration of a display device according to Embodiment 1 disclosed herein.

FIG. 2 is a perspective view schematically showing a configuration of an organic EL display unit according to Embodiment 1

FIG. 3 is a circuit diagram showing an example of a specific configuration of monitor pixel.

FIG. 4 is a block diagram showing an example of a specific configuration of a high-side potential variable-voltage source according to Embodiment 1.

FIG. **5** is a flowchart showing the operation of the display device according to Embodiment 1.

FIG. **6** is a chart showing an example of a required voltage conversion table according to Embodiment 1.

FIG. 7 is a chart showing an example of a voltage margin conversion table.

FIG. **8** is a timing chart showing the operation of the display device from an Nth frame to an N+2th frame.

 $FIG.\,9$  is diagram schematically showing images displayed on the organic EL display unit.

FIG. 10 is a block diagram showing an outline configuration of a display device according to Embodiment 2 disclosed 25 herein.

FIG. 11 is a block diagram showing an example of a specific configuration of a high-side potential variable-voltage source according to Embodiment 2.

FIG. 12 is a flowchart showing the operation of the display 30 device according to Embodiment 2.

FIG. 13 is a chart showing an example of a required voltage conversion table according to Embodiment 2.

FIG. 14 is a block diagram showing an outline configuration of a display device according to Embodiment 3 disclosed 35 herein.

FIG. **15** is a block diagram showing an example of a specific configuration of a high-side potential variable-voltage source according to Embodiment 3.

FIG. **16** is a timing chart showing the operation of the 40 display device from an Nth frame to an N+2th frame.

FIG. 17 is a block diagram showing an outline configuration of a display device according to Embodiment 4 disclosed herein.

FIG. **18** is a perspective view schematically showing a 45 configuration of an organic EL display unit according to Embodiment 4.

Claim 19A is a diagram of the circuit configuration of a pixel connected to a high-side potential monitor wire.

FIG. 19B is a diagram of the circuit configuration of a pixel 50 connected to a low-side potential monitor wire.

FIG. **20**A is a diagram showing an outline configuration of a display panel included in a display device disclosed herein.

FIG. 20B is perspective diagram schematically showing the vicinity of the periphery of the display panel included in a 55 display device disclosed herein.

FIG. 21 is a block diagram showing an outline configuration of a display device according to Embodiment 5 disclosed herein.

FIG. 22 is a diagram showing potential distributions and a 60 detecting point arrangement for the display device according to Embodiment 5.

FIG. 23 is a graph showing the pixel luminance of a normal pixel and the pixel luminance of a pixel having the monitor wire, which correspond to the gradation levels of video data. 65

FIG. **24** is a diagram schematically showing an image in which line defects occur.

4

FIG. **25** is a graph showing together current-voltage characteristics of the driving transistor and current-voltage characteristics of the organic EL element.

FIG. **26** is an external view of a thin flat-screen TV incorporating a display device disclosed herein.

#### DESCRIPTION OF EMBODIMENTS

A display device according to a general aspect of the 10 present disclosure includes: a power supplying unit configured to output at least one of a high-side output potential and a low-side output potential; a display unit in which pixels are arranged and which receives power supply from the power supplying unit; a voltage detecting unit configured to detect at least one of a high-side applied potential to be applied to at least one of the pixels in the display unit and a low-side applied potential to be applied to at least one of the pixels in the display unit which is the same as or different from the at least one of the pixels to which the high-side applied potential is to be applied; and a voltage regulating unit configured to regulate at least one of the high-side output potential and the low-side output potential outputted by the power supplying unit to set, to a predetermined potential difference, a potential difference between (i) at least one of the high-side applied potential and the low-side applied potential and (ii) a reference potential.

Accordingly, by regulating at least one of the high-side output potential of the power supplying unit and the low-side output potential of the power supplying unit in accordance with the amount of voltage drop occurring from the power supplying unit to at least one pixel, power consumption can be reduced.

Furthermore, in the display device according to an aspect of the present disclosure, the at least one pixel for which the high-side applied potential is detected and the at least one pixel for which the low-side applied potential is detected may be different pixels.

Accordingly, when the voltage drop distribution of the high-side potential power source line and the voltage drop (rise) distribution of the low-side potential power source line are different, the output potential of the power supplying unit can be regulated based on potential information from different pixels, and thus power consumption can be reduced more effectively.

Furthermore, in the display device according to an aspect of the present disclosure, at least one of (i) the at least one pixel for which the high-side applied potential is detected and (ii) the at least one pixel for which the low-side applied potential is detected may be plural in number.

Furthermore, in a display device according to an aspect of the present disclosure, the voltage regulating unit may be configured to: select at least one applied potential from among: a lowest applied potential among high-side applied potentials detected by the voltage detecting unit; and a highest applied potential among low-side applied potentials detected by the voltage detecting unit; and regulate the power supplying unit based on the selected at least one applied potential.

Accordingly, when one of the high-side potential and the low-side potential that are detected is plural in number, it is possible to select the lowest or highest potential among the plural detected potentials. Therefore, the output potential from the power supplying unit can be more accurately regulated. Therefore, power consumption can be effectively reduced even when the size of the display unit is increased.

Furthermore, a display device according to an aspect of the present disclosure may further include at least one of: a highside potential detecting line having one end connected to the

at least one pixel for which the high-side applied potential is detected and the other end connected to the voltage detecting unit, for transmitting the high-side applied potential to the voltage detecting unit; and a low-side potential detecting line having one end connected to the at least one pixel for which 5 the low-side applied potential is detected and the other end connected to the voltage detecting unit, for transmitting the low-side applied potential to the voltage detecting unit.

With this, the voltage detecting unit can measure at least one of (i) the high-side potential applied to the at least one 10 pixel via the high-side potential detecting line and (ii) the low-side potential applied to at least one pixel via the low-side potential detecting line.

Furthermore, in a display device according to an aspect of the present disclosure, the voltage detecting unit may be 15 further configured to detect at least one of the high-side output potential and the low-side output potential which are output-ted by the power supplying unit, and the voltage regulating unit may be configured to regulate at least one of the high-side output potential and the low-side output potential which are 20 outputted by the power supplying unit, in accordance with a first potential difference which is at least one of (i) a potential difference between the high-side output potential and the high-side applied potential and (ii) a potential difference between the low-side output potential and the low-side 25 applied potential.

Accordingly, since the voltage detecting unit can actually measure the voltage drop amount from the power supplying unit up to a predetermined pixel, at least one of the high-side output potential of the power supplying unit and the low-side output potential of the power supplying unit can be set to an optimal potential that is in accordance with the voltage drop amount measured by the voltage detecting unit.

Furthermore, in a display device according to an aspect of the present disclosure, the voltage regulating unit may be 35 configured to perform the regulation that realizes an increasing function relationship between (i) at least one of a potential difference between the high-side output potential and the reference potential and a potential difference between the low-side output potential and the reference potential and (ii) 40 the first potential difference.

Furthermore, in a display device according to an aspect of the present disclosure, the voltage detecting unit may be further configured to detect at least one of (i) a potential in a high-side potential current path connecting the power supply- 45 ing unit and a high potential side of the pixels and (ii) a potential in a low-side potential current path connecting the power supplying unit and the low potential side of the pixels; and the voltage regulating unit may be configured to regulate at least one of the high-side output potential and the low-side 50 output potential that are outputted by the power supplying unit, in accordance with a second potential difference which is at least one of (i) a potential difference between the potential in the high-side potential current path and the high-side applied potential and (ii) a potential difference between the 55 potential in the low-side potential current path and the lowside applied potential.

Accordingly, the output voltage from the power supplying unit can be regulated in accordance with the voltage drop amount within the display region only, by detecting the potential difference between the voltage applied to the pixels and the voltage in the wiring path outside the display region.

Furthermore, in a display device according to an aspect of the present disclosure, the voltage regulating unit may be configured to perform the regulation that realizes an increasing function relationship between (i) at least one of a potential difference between the high-side output potential and the 6

reference potential and a potential difference between the low-side output potential and the reference potential and (ii) the second potential difference.

Furthermore, in a display device according to an aspect of the present disclosure, each of the pixels may include: a driving element having a source electrode and a drain electrode; and a luminescence element having a first electrode and a second electrode, the first electrode being connected to one of the source electrode and the drain electrode of the driving element, the high-side applied potential may be applied to one of the second electrode and the other of the source electrode and the drain electrode applied potential may be applied to the other of the second electrode and the other of the second electrode and the other of the source electrode and the drain electrode.

Furthermore, in a display device according to an aspect of the present disclosure, the pixels may be arranged in rows and columns, the display device may further include a first power source line and a second power source line, the first power source line connecting the others of the source electrode and the drain electrode of the respective driving elements of adjacent pixels in at least one of the row direction and the column direction, and the second power source line connecting the second electrodes of the respective luminescence elements of adjacent pixels in the row direction and the column direction, and the pixels may receive the power supply from the power supplying unit via the first power source line and the second power source line.

Furthermore, in a display device according to an aspect of the present disclosure, the second electrode and the second power source line may be part of a common electrode that is common to the pixels, and may be electrically connected to the power supplying unit so that a potential is applied to the common electrode from a periphery of the common electrode.

Accordingly, although the voltage drop amount becomes larger towards the vicinity of the center of the display unit, the high-side output potential of the power supplying unit and the low-side output potential of the power supplying unit can be appropriately regulated and thus power consumption can be further reduced, particularly when the size of the display unit is increased.

Furthermore, in a display device according to an aspect of the present disclosure, the second electrode may comprise a transparent conductive material including a metal oxide.

Furthermore, in a display device according to an aspect of the present disclosure, the luminescence element may be an organic electroluminescence (EL) element.

Accordingly, since heat generation can be suppressed through the reduction of power consumption, the deterioration of the organic EL element can be suppressed.

These general and specific aspects may be implemented using a system, a method, an integrated circuit, a computer program, or a computer-readable recording medium such as a CD-ROM, or any combination of systems, methods, integrated circuits, computer programs, or computer-readable recording media.

Hereinafter, certain exemplary embodiments are described in greater detail with reference to the accompanying Drawings. It is to be noted that, in all the figures, the same reference numerals are given to the same or corresponding elements and redundant description thereof shall be omitted.

Each of the exemplary embodiments described below shows a general or specific example. The numerical values, shapes, materials, structural elements, the arrangement and connection of the structural elements, steps, the processing order of the steps etc. shown in the following exemplary embodiments are mere examples, and therefore do not limit

the scope of the appended Claims and their equivalents. Therefore, among the structural elements in the following exemplary embodiments, structural elements not recited in any one of the independent claims are described as arbitrary structural elements.

#### Embodiment 1

A display device according to this embodiment includes: a variable-voltage source which outputs at least one of a highside output potential and a low-side output potential; an organic electroluminescence (EL) display unit in which pixels are arranged and which receives power supply from the variable-voltage source; a potential difference detecting circuit which detects at least one of a high-side applied potential 15 to be applied to one pixel in the organic EL display unit and a low-side applied potential to be applied to at least one pixel that is the same or different from the one pixel; and a signal processing circuit which regulates at least one of the high-side output potential and the low-side output potential outputted 20 from the variable-voltage source to set, to a predetermined potential difference, a potential difference between (i) at least one of the high-side applied potential and the low-side applied potential to be applied to the pixels and (ii) a reference

Accordingly, the display device according to this embodiment realizes excellent power consumption reducing effect.

Hereinafter, Embodiment 1 shall be specifically described with reference to the Drawings.

FIG. 1 is a block diagram showing an outline configuration 30 of the display device according to Embodiment 1.

A display device **50** shown in the figure includes an organic electroluminescence (EL) display unit **110**, a data line driving circuit **120**, a write scan driving circuit **130**, a control circuit **140**, a signal processing circuit **165**, a high-side potential 35 difference detecting circuit **170**, a voltage margin setting unit **175**, a high-side potential variable-voltage source **180**, and a monitor wire **190**.

FIG. **2** is a perspective view schematically showing a configuration of the organic EL display unit **110** according to 40 Embodiment 1. It is to be noted that the lower portion of the figure is the display screen side.

As shown in the figure, the organic EL display unit 110 includes the pixels 111, the first power source wire 112, and the second power source wire 113.

Each pixel 111 is connected to the first power source wire 112 and the second power source wire 113, and produces luminescence at a luminance that is in accordance with a pixel current ipix that flows to the pixel 111. At least one predetermined pixel out of the pixels 111 is connected to the monitor 50 wire 190 at a high-side potential detecting point M1. Hereinafter, the pixel 111 that is directly connected to the monitor wire 190 shall be denoted as monitor pixel 111M.

The first power source wire 112 is arranged in a net-like manner to correspond to pixels 111 that are arranged in a 55 matrix, and is electrically connected to the high-side potential variable-voltage source 180 disposed at the periphery of the organic EL display unit 110. Through the outputting of a high-side power source potential from the high-side potential variable-voltage source 180, a potential corresponding to the high-side power source potential outputted by the high-side potential variable-voltage source 180 is applied to the first power source wire 112. On the other hand, the second power source wire 113 is formed in the form of a continuous film on the organic EL display unit 110. In FIG. 2, the first power 65 source wire 112 and the second power source wire 113 are schematically illustrated in mesh-form in order to show the

8

resistance components of the first power source wire 112 and the second power source wire 113. It is to be noted that the second power source wire 113 may be grounded to, for example, a common grounding potential of the display device 50, at the periphery of the organic EL display unit 110.

A horizontal first power source wire resistance  $R1\nu$  are present in the first power source wire 112. A horizontal second power source wire resistance  $R2\nu$  are present in the first power source wire 112. A horizontal second power source wire resistance  $R2\nu$  and a vertical second power source wire resistance  $R2\nu$  are present in the second power source wire 113. It is to be noted that, although not illustrated, each of the pixels 111 is connected to the write scan driving circuit 130 and the data line driving circuit 120, and is also connected to a scanning line for controlling the timing at which the pixel produces luminescence and stops producing luminescence, and to a data line for supplying signal voltage corresponding to the pixel luminance of the pixel 111.

The optimal position of the monitor pixel 111M is determined depending on the wiring method of the first power source wire 112, and the respective values of the first power source wire resistances R1h and R1v.

FIG. 3 is a circuit diagram showing an example of a specific configuration of the monitor pixel 111M. Each of the pixels arranged in a matrix includes a driving element and a luminescence element. The driving element includes a source electrode and a drain electrode. The luminescence element includes a first electrode and a second electrode. The first electrode is connected to one of the source electrode and the drain electrode of the driving element. The high-side potential is applied to one of (i) the other of the source electrode and the drain electrode and (ii) the second electrode, and the low-side potential is applied to the other of (i) the other of the source electrode and the drain electrode and the drain electrode and (ii) the second electrode

Specifically, each of the pixels 111 includes an organic EL element 121, a data line 122, a scanning line 123, a switch transistor 124, a driving transistor 125, and a holding capacitor 126. Furthermore, in the monitor pixel 111M, the monitor wire 190A is additionally connected to the other of the source electrode and the drain electrode of the driving element. At least one pixel 111M is disposed in the organic EL display unit 110.

The organic EL element 121 is a luminescence element having an anode electrode, which is a first electrode, connected to the drain electrode of the driving transistor 125 and a cathode electrode, which is a second electrode, connected to the second power source wire 113, and produces luminescence with a luminance that is in accordance with the pixel current ipix flowing between the anode electrode and the cathode electrode. The cathode electrode of the organic EL element 121 forms part of a common electrode provided in common to the plural pixels 111, and potential is applied to the common electrode from the periphery of the common electrode. Specifically, the common electrode functions as the second power source wire 113 in the organic EL display unit 110. Furthermore, the cathode electrode is formed from a transparent conductive material made of a metallic oxide.

The data line 122 is connected to the data line driving circuit 120 and one of the source electrode and the drain electrode of the switch transistor 124, and signal voltage corresponding to video data is applied to the data line 122 by the data line driving circuit 120.

The scanning line 123 is connected to the write scan driving circuit 130 and the gate electrode of the switch transistor 124, and switches between conduction and non-conduction of the switching transistor 124 according to the voltage applied by the write scan driving circuit 130.

The switching transistor 124 has one of a source electrode and a drain electrode connected to the data line 122, the other of the source electrode and the drain electrode connected to the gate electrode of the driving transistor 125 and one end of the holding capacitor 126, and is, for example, a P-type thin-5 film transistor (TFT).

The driving transistor 125 is a driving element having a source electrode connected to first power source wire 112, a drain electrode connected to the anode electrode of the organic EL element 121, and a gate electrode connected to the 10 one end of the holding capacitor 126 and the other of the source electrode and the drain electrode of the switching transistor 124, and is, for example, a P-type TFT. With this, the driving transistor 125 supplies the organic EL element 121 with current that is in accordance with the voltage held in 15 the holding capacitor 126. Furthermore, in the monitor pixel 111M, the source of the driving transistor 125 is connected to the monitor wire 190.

The holding capacitor 126 has one end connected to the other of the source electrode and the drain electrode of the 20 switch transistor 124, and the other end connected to the first power source wire 112, and holds the potential difference between the potential of the first power source wire 112 and the potential of the gate electrode of the driving transistor 125 when the switch transistor 124 becomes non-conductive. 25 Specifically, the holding capacitor 126 holds a voltage corresponding to the signal voltage.

The functions of the respective constituent elements shown in FIG. 1 shall be described below with reference to FIG. 2 and FIG. 3.

The data line driving circuit 120 outputs signal voltage corresponding to video data, to the pixels 111 via the data lines 122.

The write scan driving circuit 130 sequentially scans the pixels 111 by outputting a scanning signal to scanning lines 35 given.

123. Specifically, the switch transistors 124 are switched between conduction and non-conduction on a per row basis. With this, the signal voltages outputted to the data lines 122 accord are applied to the pixels 111 in the row selected by the write scan driving circuit 130. Therefore, the pixels 111 produce 40 which luminescence with a luminance that is in accordance with the video data.

The control circuit 140 instructs the drive timing to each of the data line driving circuit 120 and the write scan driving circuit 130.

The signal processing circuit **165** outputs, to the data line driving circuit **120**, a signal voltage corresponding to inputted video data

The high-side potential difference detecting circuit 170, which is the voltage detecting unit in this embodiment, mea- 50 sures, for the monitor pixel 111M, the high-side potential to be applied to the monitor pixel 111M. Specifically, the highside potential difference detecting circuit 170 measures, via the monitor wire 190, the high-side potential to be applied to the monitor pixel 111M. Specifically, the high-side potential 55 difference detecting circuit 170 measures the potential at the detecting point M1. In addition, the high-side potential difference detecting circuit 170 measures the output potential of the high-side potential variable-voltage source 180, and measures the potential difference  $\Delta V$  between the measured highside potential to be applied to the monitor pixel 111M and the output potential of the high-side potential variable-voltage source 180. Subsequently, the high-side potential difference detecting circuit 170 outputs the measured potential difference  $\Delta V$  to the voltage margin setting unit 175.

The voltage margin setting unit 175, which is the voltage regulating unit in this embodiment, regulates, based on a

10

voltage (VEL+VTFT) at a peak gradation level and the potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170, the high-side potential variable-voltage source 180 to set the potential difference between the potential of the monitor pixel 111M and the reference potential to a predetermined voltage. Specifically, the voltage margin setting unit 175 calculates a voltage drop margin Vdrop, based on the potential difference detected by the high-side potential difference detecting circuit 170. Subsequently, the voltage margin setting unit 175 sums up the voltage (VEL+VTFT) at the peak gradation level and the voltage drop margin Vdrop, and outputs the summation result VEL+VTFT+ Vdrop, as the voltage of a first reference voltage Vref1A, to the high-side potential variable-voltage source 180.

The high-side potential variable-voltage source 180, which is the power supplying unit in this embodiment, outputs the high-side potential to the organic EL display unit 110. The high-side potential variable-voltage source 180 outputs an output voltage Vout for setting the potential difference between the high-side potential of the monitor pixel 111M and the reference potential to the predetermined voltage (VEL+VTFT), according to the first reference voltage Vref1A outputted by the voltage margin setting unit 175. It is sufficient that reference potential be a potential serving as a reference in the display device 50, and is for example a ground potential in this embodiment.

The monitor wire 190 is a high-side potential detecting line which has one end connected to the monitor pixel 111M and the other end connected to the high-side potential difference detecting circuit 170, and transmits the high-side potential applied to the monitor pixel 111M to the high-side potential difference detecting circuit 170.

Next, a brief description of the detailed configuration of the high-side potential variable-voltage source 180 shall be given.

FIG. 4 is a block diagram showing an example of a specific configuration of a high-side potential variable-voltage source according to Embodiment 1. It is to be noted that the organic EL display unit 110 and the voltage margin setting unit 175 which are connected to the variable-voltage source are also shown in the figure.

The high-side potential variable-voltage source 180 shown in the figure includes a comparison circuit 181, a pulse width modulation (PWM) circuit 182, a drive circuit 183, a switching element SW, a diode D, an inductor L, a capacitor C, and an output terminal 184, and converts an input voltage Vin into an output voltage Vout which is in accordance with the first reference voltage Vref1, and outputs the output voltage Vout from the output terminal 184. It is to be noted that, although not illustrated, an AC-DC converter is provided in a stage ahead of an input terminal to which the input voltage Vin is inputted, and it is assumed that conversion, for example, from 100V AC to 20V DC is already carried out.

The comparison circuit 181 includes an output detecting unit 185 and an error amplifier 186, and outputs a voltage that is in accordance with the difference between the output voltage Vout and the first reference voltage Vref1A, to the PWM circuit 182.

The output detecting unit 185, which includes two resistors R1 and R2 provided between the output terminal 184 and a grounding potential, voltage-divides the output voltage Vout in accordance with the resistance ratio between the resistors R1 and R2, and outputs the voltage-divided output voltage Vout to the error amplifier 186.

The error amplifier **186** compares the Vout that has been voltage-divided by the output detection unit **185** and the first reference voltage Vref1A outputted by the voltage margin

setting unit 175, and outputs, to the PWM circuit 182, a voltage that is in accordance with the comparison result. Specifically, the error amplifier 186 includes an operational amplifier 187 and resistors R3 and R4. The operational amplifier 187 has an inverting input terminal connected to the 5 output detecting unit 185 via the resistor R3, a non-inverting input terminal connected to the voltage margin setting unit 175, and an output terminal connected to the PWM circuit **182**. Furthermore, the output terminal of the operational amplifier 187 is connected to the inverting input terminal via the resistor R4. With this, the error amplifier 186 outputs, to the PWM circuit 182, a voltage that is in accordance with the potential difference between the voltage inputted from the output detecting unit 185 and the first reference voltage Vref1A inputted from the voltage margin setting unit 175. Stated differently, the error amplifier 186 outputs, to the PWM circuit 182, a voltage that is in accordance with the potential difference between the output voltage Vout and the first reference voltage Vref1A.

The PWM circuit **182** outputs, to the drive circuit **183**, 20 pulse waveforms having different duties depending on the voltage outputted by the comparison circuit **181**. Specifically, the PWM circuit **182** outputs a pulse waveform having a long ON duty when the voltage outputted by the comparison circuit **181** is large, and outputs a pulse waveform having a short 25 ON duty when the outputted voltage is small. Stated differently, the PWM circuit **182** outputs a pulse waveform having a long ON duty when the potential difference between the output voltage Vout and the first reference voltage Vref**1**A is big, and outputs a pulse waveform having a short ON duty when the potential difference between the output voltage Vout and the first reference voltage Vref**1**A is small. It is to be noted that the ON period of a pulse waveform is a period in which the pulse waveform is active.

The drive circuit **183** turns ON the switch SW during the period in which the pulse waveform outputted by the PWM circuit **182** is active, and turns OFF the switch SW during the period in which the pulse waveform outputted by the PWM circuit **182** is inactive.

The switching element SW is switched between conduction and non-conduction by the drive circuit **183**. The input voltage Vin is outputted, as the output voltage Vout, to the output terminal **184** via the inductor L and the capacitor C only while the switching element is in the state of conduction. Accordingly, from 0V, the output voltage Vout gradually 45 approaches 20V (Vin). At this time the inductor L and the capacitor C are charged. Since voltage is applied (charged) to both ends of the inductor L, the output voltage Vout becomes a potential which is lower than the input voltage Vin by such voltage.

As the output voltage Vout approaches the first reference voltage Vref1A, the voltage inputted to the PWM circuit 182 becomes smaller, and the ON duty of the pulse signal outputted by the PWM circuit 182 becomes shorter.

Then, the time in which the switching element SW is ON 55 also becomes shorter, and the output voltage Vout gently converges with the first reference voltage Vref1A.

The potential of the output voltage Vout, while having slight voltage fluctuations, eventually settles to a potential in the vicinity of Vout=Vref1A.

In this manner, the high-side potential variable-voltage source 180 generates the output voltage Vout which becomes the first reference voltage Vref1A outputted by the signal processing circuit 165, and supplies the output voltage Vout to the organic EL display unit 110.

Next, the operation of the aforementioned display device **50** shall be described using FIG. **5** to FIG. **7**.

12

FIG. **5** is a flowchart showing the operation of the display device **50** according to Embodiment 1.

First, the voltage margin setting unit 175 reads, from a memory, the preset voltage (VEL+VTFT) corresponding to the peak gradation level (step S10). Specifically, voltage margin setting unit 175 determines the VTFT+VEL corresponding to the gradation levels for each color, using a required voltage conversion table indicating the required voltage VTFT+VEL corresponding to the gradation levels for each color.

FIG. 6 is a chart showing an example of a required voltage conversion table which is referenced by the voltage margin setting unit 175. As shown in the figure, required voltages VTFT+VEL respectively corresponding to the peak gradation level (gradation level 255) are stored in the required voltage at the peak gradation level of R is 11.2 V, the required voltage at the peak gradation level of G is 12.2 V, and the required voltage at the peak gradation level of B is 8.4 V. Among the required voltages at the peak gradation levels of the respective colors, the largest voltage is the 12.2 V of G. Therefore, the voltage margin setting unit 175 determines VTFT+VEL to be 12.2 V.

Meanwhile, the high-side potential difference detecting circuit 170 detects the potential at the detecting point M1 via the monitor wire 190 (step S14).

Next, the high-side potential difference detecting circuit 170 receives the voltage of the output terminal 184 of the high-side potential variable-voltage source 180, calculates the potential difference between potential at the detecting point M1 and the reference potential, and detects the potential difference  $\Delta V$  between the received potential of the output terminal 184 and the calculated potential difference (step S15). Subsequently, the high-side potential difference detecting circuit 170 outputs the detected potential difference  $\Delta V$  to the voltage margin setting unit 175. It is to be noted that the steps S10 to S15 up to this point correspond to the potential measuring process according to the present disclosure.

Next, the voltage margin setting unit 175 determines a voltage drop margin Vdrop corresponding to the potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170, based on a potential difference signal outputted by the high-side potential difference detecting circuit 170 (step S16). Specifically, the voltage margin setting unit 175 has a voltage margin conversion table indicating the voltage drop margin Vdrop corresponding to the potential difference  $\Delta V$ , and determines the voltage drop margin Vdrop with reference to the conversion table.

FIG. 7 is a chart showing an example of the voltage margin conversion table that is referenced by the voltage margin setting unit 175.

As shown in the figure, voltage drop margins  $Vdrop\ respectively$  corresponding to the potential differences  $\Delta V$  are stored in the voltage margin conversion table. For example, when the potential difference  $\Delta V$  is 3.4 V, the voltage drop margin Vdrop is 3.4 V. Therefore, the voltage margin setting unit 175 determines the voltage drop margin Vdrop to be 3.4 V

Now, as shown in the voltage margin conversion table, the potential difference  $\Delta V$  and the voltage drop margin Vdrop have an increasing function relationship. Furthermore, the output voltage Vout of the high-side potential variable-voltage source  $180\,\mathrm{rises}$  with a bigger voltage drop margin Vdrop. In other words, the potential difference  $\Delta V$  and the output voltage Vout have an increasing function relationship.

Next, the voltage margin setting unit 175 determines the output voltage Vout that the high-side potential variable-volt-

age source 180 is to be made to output in the next frame period (step S17). Specifically, the output voltage Vout that the highside potential variable-voltage source 180 is to be made to output in the next frame period is assumed to be VTFT+VEL+ Vdrop which is the sum value of (i) VTFT+VEL determined 5 in the determination (step S13) of the voltage required by the organic EL element 121 and the driving transistor 125 and (ii) the voltage drop margin Vdrop determined in the determination (step S15) of the voltage margin corresponding to the potential difference  $\Delta V$ .

Lastly, the voltage margin setting unit 175 regulates the high-side potential variable-voltage source 180 by setting the first reference voltage Vref1A as VTFT+VEL+Vdrop at the beginning of the next frame period (step S18). With this, in the next frame period, the high-side potential variable-voltage 15 source 180 supplies Vout=VTFT+VEL+Vdrop to the organic EL display unit 110. It is to be noted that step S16 to step S18 correspond to the voltage regulating process according to the present disclosure.

In this manner, the display device 50 according to this 20 embodiment includes: the high-side potential variable-voltage source 180 which outputs the high-side potential; the high-side potential difference detecting circuit 170 which measures, for the monitor pixel 111M in the organic EL display unit 110, (i) the high-side potential applied to the 25 monitor pixel 111M and (ii) the high-side potential output voltage Vout of the high-side potential variable-voltage source 180; and the voltage margin setting unit 175 which regulates the high-side potential variable-voltage source 180 to set, to the predetermined potential (VTFT+VEL), the 30 potential difference between the reference potential and the high-side potential that is applied to the monitor pixel 111M that is measured by the high-side potential difference detecting circuit 170. Furthermore, the high-side potential difference detecting circuit 170, in addition, measures the high-side 35 potential output voltage Vout of the high-side potential variable-voltage source 180, detects the potential difference between the measured high-side potential output voltage Vout and the high-side potential applied to the monitor pixel 111M. The voltage margin setting unit 175 regulates the high-side 40 potential variable-voltage source 180 in accordance with the potential difference detected by the high-side potential difference detecting circuit 170.

With this, the display device **50** can reduce excess voltage and reduce power consumption by detecting the voltage drop 45 caused by the horizontal first power source wire resistance R1h and a vertical first power source wire resistance R1v and giving feedback to the high-side potential variable-voltage source **180** regarding the degree of such voltage drop.

Furthermore, in the display device **50**, the monitor pixel 50 **111M** is located near the center of the organic EL display unit **110**, and thus the output voltage Vout of the high-side potential variable-voltage source **180** can be easily regulated even when the size of the organic EL display unit **110** is increased.

Furthermore, since heat generation by the organic EL ele- 55 ment **121** is suppressed through the reduction of power consumption, the deterioration of the organic EL element **121** can be prevented.

Next, the display pattern transition in the case where the video data inputted up to the Nth frame changes from the 60 N+1th frame onward, in the display device 50 described above, shall be described using FIG. 8 and FIG. 9.

Initially, the video data that is assumed to have been inputted in the Nth frame and the N+1th frame shall be described.

First, it is assumed that, up to the Nth frame, the video data 65 corresponding to the central part of the organic EL display unit 110 is a peak gradation level (R:G:B=255:255:255) in

14

which the central part of the organic EL display unit 110 is seen as being white. On the other hand, it is assumed that the video data corresponding to a part of the organic EL display unit 110 other than the central part is a gray gradation level (R:G:B=50:50:50) in which the part of the organic EL display unit 110 other than the central part is seen as being gray.

Furthermore, from the N+1th frame onward, it is assumed that the video data corresponding to the central part of the organic EL display unit **110** is the peak gradation level (R:G: B=255:255:255) as in the Nth frame. On the other hand, it is assumed that the video data corresponding to the part of the organic EL display unit **110** other than the central part is a gray gradation level (R:G:B=150:150:150) that can be seen as a brighter gray than in the Nth frame.

Next, the operation of the display device **50** in the case where video data as described above is inputted in the Nth frame and the N+1th frame shall be described.

FIG. **8** is a timing chart showing the operation of the display device **50** from the Nth frame to the N+2th frame.

The potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170, the output voltage Vout from the high-side potential variable-voltage source 180, and the pixel luminance of the monitor pixel 111M are shown in the figure. Furthermore, a blanking period is provided at the end of each frame period.

FIG. 9 is diagram schematically showing images displayed on the organic EL display unit.

In a  $t=\bar{T}10$ , the signal processing circuit 165 receives input of the video data of the Nth frame. The voltage margin setting unit 175 uses the required voltage conversion table and sets the  $12.2\,\mathrm{V}$  required voltage at the peak gradation level of G to the voltage (VTFT+VEL).

Meanwhile, at this time, the high-side potential difference detecting circuit 170 detects the potential at the detecting point M1 via the monitor wire 190, and detects the potential difference  $\Delta V$  between the detected potential and the output voltage Vout outputted from the high-side potential variable-voltage source 180. For example, in time t=T10, the high-side potential difference detecting circuit 170 detects  $\Delta V{=}1V.$  Subsequently, the voltage margin setting unit 175 uses the voltage margin conversion table and determines the voltage drop margin Vdrop for the N+1th frame to be 1 V.

A time t=T10 to T11 is the blanking period of the Nth frame. In this period, an image which is the same as that in the time t=T10 is displayed in the organic EL display unit 110.

(a) in FIG. 9 schematically shows an image displayed on the organic EL display unit 110 in the time t=T10 to T11. In this period, the image displayed on the organic EL display unit 110 corresponds to the image data of the Nth frame, and thus the central part is white and the part other than the central part is gray.

In time t=T11, the voltage margin setting unit 175 sets the voltage of the first reference voltage Vref1A as the sum VTFT+VEL+Vdrop (for example, 13.2 V) of the aforementioned voltage (VTFT+VEL) and the voltage drop margin Vdrop.

Over a time t=T11 to T16, the image corresponding to the video data of the N+1th frame is gradually displayed on the organic EL display unit 110 ((b) to (f) in FIG. 9). At this time, the output voltage Vout from the high-side potential variable-voltage source 180 is, at all times, the VTFT+VEL+Vdrop set to the voltage of the first reference voltage Vref1A in time t=T11. However, the video data corresponding to the part of the organic EL display unit 110 other than the central part is a gray gradation level that can be seen as a gray that is brighter than that in the Nth frame. Therefore, the amount of current supplied by the high-side potential variable-voltage source

180 to the organic EL display unit 110 gradually increases over time t=T11 to T16, and the voltage drop in the first power source wire 112 gradually increases following this increase in the amount of current. With this, there is a shortage of power source voltage for the pixels 111 in the central part of the 5 organic EL display unit 110, which are the pixels 111 in a brightly displayed region. Stated differently, luminance drops below the image corresponding to the video data R:G:B=255: 255:255:255 of the N+1th frame. Specifically, over the time t=T11 to T16, the pixel luminance of the pixels 111 at the central part 10 of the organic EL display unit 110 gradually drops.

In a time t=T16, the signal processing circuit 165 receives input of the video data of the N+1th frame. The voltage margin setting unit 175 uses the required voltage conversion table and continues to set the 12.2 V required voltage at the 15 peak gradation level of G to the voltage (VTFT+VEL).

Meanwhile, at this time, the high-side potential difference detecting circuit 170 detects the potential at the detecting point M1 via the monitor wire 190, and detects the potential difference  $\Delta V$  between (i) the potential difference between 20 the detected potential at the detecting point M1 and the reference potential and (ii) and the output voltage Vout outputted from the high-side potential variable-voltage source 180. For example, in time t=T16, the high-side potential difference detecting circuit 170 detects  $\Delta V{=}3$  V. Subsequently, the voltage margin setting unit 175 uses the voltage margin conversion table and determines the voltage drop margin Vdrop for the N+1th frame to be 3 V.

Next, in time t=T17, the voltage margin setting unit 175 sets the voltage of the first reference voltage Vref1A to the 30 sum VTFT+VEL+Vdrop (for example, 15.2 V) of the aforementioned voltage (VTFT+VEL) and the voltage drop margin Vdrop. Therefore, from time t=T17 onward, the potential at the detecting point M1 becomes VTFT+VEL which is the predetermined potential.

In this manner, in the display device 100, although luminance temporarily drops in the N+1th frame, this is a very short period and thus has practically no impact on the user.

It is to be noted that although, in this embodiment, the high-side potential applied to the pixel 111M is monitored 40 and the high-side potential outputted from the high-side potential variable-voltage source 180 is regulated to set the potential difference between the monitored high-side potential and the reference potential to a predetermined voltage, the low-side potential applied to the pixel 111M may be moni- 45 tored and the low-side potential outputted from a low-side potential variable-voltage source may be regulated to set the potential difference between the monitored low-side potential and the reference potential to a predetermined voltage. In this case, a low-side potential difference detecting circuit is 50 provided in place of the high-side potential difference detecting circuit 170 shown in FIG. 1, and a low-side potential variable-voltage source is provided in place of the high-side potential variable-voltage source 180.

The function of the voltage margin setting unit is the same 55 as that of the voltage margin setting unit 175. The voltage margin setting unit regulates, based on a voltage (VEL+ VTFT) at a peak gradation level and the potential difference  $\Delta V$  detected by the low-side potential difference detecting circuit, the low-side potential variable-voltage source to set 60 the potential of the monitor pixel 111M to a predetermined potential.

The low-side potential difference detecting circuit measures, for the monitor pixel 111M, a low-side potential applied to the monitor pixel 111M. In addition, the low-side 65 potential difference detecting circuit measures the output potential of the low-side potential variable-voltage source,

16

and measures the potential difference  $\Delta V$  between (i) the potential difference between the measured low-side potential applied to the monitor pixel 111M and the reference potential and (ii) the output potential of the low-side variable-voltage source. Subsequently, the low-side potential difference detecting circuit outputs the measured potential difference  $\Delta V$  to the voltage margin setting unit.

The low-side potential variable-voltage source outputs the low-side potential to the organic EL display unit 110. The low-side potential variable-voltage source outputs an output voltage Vout for setting the potential difference between the low-side potential of the monitor pixel 111M and the reference potential to a predetermined potential, according to a reference voltage Vref outputted by the voltage margin setting unit. It is sufficient that reference potential be a potential serving as a reference in the display device 50.

With this, the display device in the present disclosure can reduce excess voltage and reduce power consumption by detecting the voltage drop caused by the horizontal second power source wire resistance R2h and the vertical second power source wire resistance R2v and giving feedback to the low-side potential variable-voltage source regarding the degree of such voltage drop.

Furthermore, since heat generation by the organic EL element is suppressed through the reduction of power consumption, the deterioration of the organic EL element can be prevented.

#### Embodiment 2

Compared to the display device according to Embodiment 1, a display device according to this embodiment is different in that the reference voltage that is inputted to a variable-voltage source not only changes depending on a change in the potential difference  $\Delta V$  detected by a high-side potential difference detecting circuit, but also changes depending on a peak signal detected, for each frame, from the inputted video data. Hereinafter, description shall not be repeated for points which are the same as in Embodiment 1 and shall be centered on the points of difference from Embodiment 1. Furthermore, the figures applied to Embodiment 1 shall be used for figures that would otherwise overlap with those in Embodiment 1.

Hereinafter, Embodiment 2 shall be specifically described with reference to the Drawings.

FIG. 10 is a block diagram showing an outline configuration of the display device according to Embodiment 2.

A display device 100 shown in the figure includes the organic electroluminescence (EL) display unit 110, the data line driving circuit 120, the write scan driving circuit 130, the control circuit 140, a peak signal detecting circuit 150, a signal processing circuit 160, the high-side potential difference detecting circuit 170, the high-side potential variable-voltage source 180, and the monitor wire 190.

The configuration of the organic EL display unit 110 is the same as that shown in FIG. 2 and FIG. 3 in Embodiment 1.

The peak signal detecting circuit 150 detects the peak value of the video data inputted to the display device 100, and outputs a peak signal representing the detected peak value to the signal processing circuit 160. Specifically, the peak signal detecting circuit 150 detects, as the peak value, data of the highest gradation level out of the video data. High gradation level data corresponds to an image that is to be displayed brightly by the organic EL display unit 110.

The signal processing circuit 160, which is the voltage regulating unit in this embodiment, regulates the variable-voltage source 180 to set the potential difference between the potential of the monitor pixel 111M and the reference poten-

tial to a predetermined voltage, based on the peak signal outputted by the peak signal detecting circuit 150 and a potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170. Specifically, the signal processing circuit 160 determines the voltage required by the organic EL element 121 and the driving transistor 125 when causing the pixels 111 to produce luminescence according to the peak signal outputted by the peak signal detecting circuit 150. Furthermore, the signal processing circuit 160 calculates a voltage margin based on the potential difference detected by the high-side potential difference detecting circuit 170. Subsequently, the signal processing circuit 160 sums up a voltage VEL required by the organic EL element 121, a voltage VTFT required by the driving transistor 125, and the voltage drop margin Vdrop, and outputs the summation result VEL+ VTFT+Vdrop, as the voltage of a first reference voltage Vref1, to the high-side potential variable-voltage source 180.

Furthermore, the signal processing circuit 160 outputs, to the data line driving circuit 120, a signal voltage corresponding to the video data inputted via the peak signal detecting circuit 150.

The high-side potential difference detecting circuit 170, which is the voltage detecting unit in this embodiment, measures, for the monitor pixel 111M, the high-side potential to 25 be applied to the monitor pixel 111M. Specifically, the highside potential difference detecting circuit 170 measures, via the monitor wire 190, the high-side potential to be applied to the monitor pixel 111M. Specifically, the high-side potential difference detecting circuit 170 measures the potential at the detecting point M1. In addition, the high-side potential difference detecting circuit 170 measures the output potential of the high-side potential variable-voltage source 180, and measures the potential difference  $\Delta V$  between the measured highside potential to be applied to the monitor pixel 111M and the output potential of the high-side potential variable-voltage source 180. Subsequently, the high-side potential difference detecting circuit 170 outputs the measured potential difference  $\Delta V$  to the signal processing circuit 160.

The high-side potential variable-voltage source 180, which is the power supplying unit in this embodiment, outputs the high-side potential to the organic EL display unit 110. The high-side potential variable-voltage source 180 outputs an output voltage Vout for setting the potential difference 45 between the high-side potential of the monitor pixel 111M and the reference potential to the predetermined potential (VEL+VTFT), according to the first reference voltage Vref1 outputted by the signal processing circuit 160. It is sufficient that reference potential be a potential serving as a reference in 50 the display device 100, and is for example a ground potential in this embodiment.

The monitor wire **190** is a high-side potential detecting line which has one end connected to the monitor pixel **111M** and the other end connected to the high-side potential difference 55 detecting circuit **170**, and transmits the high-side potential applied to the monitor pixel **111M** to the high-side potential difference detecting circuit **170**.

Next, a brief description of the detailed configuration of the high-side potential variable-voltage source **180** shall be 60 given.

FIG. 11 is a block diagram showing an example of a specific configuration of a high-side potential variable-voltage source according to Embodiment 2. It is to be noted that the organic EL display unit 110 and the signal processing circuit 65 160 which are connected to the variable-voltage source are also shown in the figure. The high-side potential variable-

18

voltage source 180 shown in the figure is the same as the high-side potential variable-voltage source 180 described in Embodiment 1.

The error amplifier 186 compares the Vout that has been voltage-divided by the output detection unit 185 and the first reference voltage Vref1 outputted by the signal processing circuit 160, and outputs, to the PWM circuit 182, a voltage that is in accordance with the comparison result. Specifically, the error amplifier 186 includes the operational amplifier 187 and the resistors R3 and R4. The operational amplifier 187 has an inverting input terminal connected to the output detecting unit 185 via the resistor R3, a non-inverting input terminal connected to the signal processing circuit 160, and an output terminal connected to the PWM circuit 182. Furthermore, the output terminal of the operational amplifier 187 is connected to the inverting input terminal via the resistor R4. With this, the error amplifier 186 outputs, to the PWM circuit 182, a voltage that is in accordance with the potential difference between the voltage inputted from the output detecting unit 185 and the first reference voltage Vref1 inputted from the signal processing circuit 160. Stated differently, the error amplifier 186 outputs, to the PWM circuit 182, a voltage that is in accordance with the potential difference between the output voltage Vout and the first reference voltage Vref1.

The PWM circuit 182 outputs, to the drive circuit 183, pulse waveforms having different duties depending on the voltage outputted by the comparison circuit 181. Specifically, the PWM circuit 182 outputs a pulse waveform having a long ON duty when the voltage outputted by the comparison circuit 181 is large, and outputs a pulse waveform having a short ON duty when the outputted voltage is small. Stated differently, the PWM circuit 182 outputs a pulse waveform having a long ON duty when the potential difference between the output voltage Vout and the first reference voltage Vref1 is big, and outputs a pulse waveform having a short ON duty when the potential difference between the output voltage Vout and the first reference voltage Vref1 is small. It is to be noted that the ON period of a pulse waveform is a period in which the pulse waveform is active.

As the output voltage Vout approaches the first reference voltage Vref1, the voltage inputted to the PWM circuit 182 decreases, and the ON duty of the pulse signal outputted by the PWM circuit 182 becomes shorter.

Then, the time in which the switching element SW is ON becomes shorter, and the output voltage Vout gently converges with the first reference voltage Vref1.

The potential of the output voltage Vout, while having slight voltage fluctuations, eventually settles to a potential in the vicinity of Vout=Vref1.

In this manner, the high-side potential variable-voltage source 180 generates the output voltage Vout which becomes the first reference voltage Vref1 outputted by the signal processing circuit 160, and supplies the output voltage Vout to the organic EL display unit 110.

Next, the operation of the aforementioned display device 100 shall be described using FIG. 12, FIG. 13, and FIG. 7.

FIG. 12 is a flowchart showing the operation of the display device 100 disclosed herein.

First, the peak signal detecting circuit 150 obtains the video data for one frame period inputted to the display device 100 (step S11). For example, the peak signal detecting circuit 150 includes a buffer and stores the video data for one frame period in such buffer.

Next, the peak signal detecting circuit 150 detects the peak value of the obtained video data (step S12), and outputs a peak signal representing the detected peak value to the signal processing circuit 160. Specifically, the peak signal detecting

circuit **150** detects the peak value of the video data for each color. For example, for each of red (R), green (G), and blue (B), the video data is expressed using the 256 gradation levels from 0 to 255 (luminance being higher with a larger value). Here, when part of the video data of the organic EL display 5 unit **110** has R:G:B=177:124:135, another part of the video data of the organic EL display unit **110** has R:G:B=24:177: 50, and yet another part of the video data of the organic EL display unit **110** has R:G:B=10:70:176, the peak signal detecting circuit **150** detects **177** as the peak value of R, **177** 10 for the peak value of G, and **176** as the peak value of B, and outputs, to the signal processing circuit **160**, a peak signal representing the detected peak value of each color.

Next, the signal processing circuit 160 determines the voltage VTFT required by the driving transistor 125 and the 15 voltage VEL required by the organic EL element 121 when causing the organic EL element 121 to produce luminescence according to the peak values outputted by the peak signal detecting circuit 150 (step S13). Specifically, the signal processing circuit 160 determines the VTFT+VEL corresponding to the gradation levels for each color, using a required voltage conversion table indicating the required voltage VTFT+VEL corresponding to the gradation levels for each color.

FIG. 13 is a chart showing an example of the required 25 voltage conversion table provided in the signal processing circuit 160.

As shown in the figure, required voltages VTFT+VEL respectively corresponding to the gradation levels of each color are stored in the required voltage conversion table. For 30 example, the required voltage corresponding to the peak value 177 of R is 8.5V, the required voltage corresponding to the peak value 177 of G is 9.9V, and the required voltage corresponding to the peak value 176 of B is 6.7V. Among the required voltages corresponding to the peak values of the 35 respective colors, the largest voltage is 9.9V corresponding to the peak value of G. Therefore, the signal processing circuit 160 determines VTFT+VEL to be 9.9V.

Meanwhile, the high-side potential difference detecting circuit 170 detects the potential at the detecting point M1 via 40 the monitor wire 190 (step S14).

Next, the high-side potential difference detecting circuit 170 receives the voltage of the output terminal 184 of the high-side potential variable-voltage source 180, calculates the potential difference between potential at the detecting 45 point M1 and the reference potential, and detects the potential difference  $\Delta V$  between the received potential of the output terminal 184 and the calculated potential difference (step S15). Subsequently, the high-side potential difference detecting circuit 170 outputs the detected potential difference  $\Delta V$  to 50 the signal processing circuit 160. It is to be noted that the steps S11 to S15 up to this point correspond to the potential measuring process according to the present disclosure.

Next, the signal processing circuit 160 determines a voltage drop margin Vdrop corresponding to the potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170, based on a potential difference signal outputted by the high-side potential difference detecting circuit 170 (step S16). Specifically, the signal processing circuit 160 has a voltage margin conversion table indicating the 60 voltage drop margin Vdrop corresponding to the potential difference  $\Delta V$ , and determines the voltage drop margin Vdrop with reference to the conversion table.

As shown in FIG. 7, voltage drop margins Vdrop respectively corresponding to the potential differences  $\Delta V$  are 65 stored in the voltage margin conversion table. For example, when the potential difference  $\Delta V$  is 3.4 V, the voltage drop

20

margin Vdrop is 3.4 V. Therefore, the signal processing circuit **160** determines the voltage drop margin Vdrop to be 3.4 V.

Now, as shown in the voltage margin conversion table, the potential difference  $\Delta V$  and the voltage drop margin Vdrop have an increasing function relationship. Furthermore, the output voltage Vout of the high-side potential variable-voltage source 180 rises with a bigger voltage drop margin Vdrop. In other words, the potential difference  $\Delta V$  and the output voltage Vout have an increasing function relationship.

Next, the signal processing circuit 160 determines the output voltage Vout that the high-side potential variable-voltage source 180 is to be made to output in the next frame period (step S17). Specifically, the output voltage Vout that the high-side potential variable-voltage source 180 is to be made to output in the next frame period is assumed to be VTFT+VEL+ Vdrop which is the sum value of (i) VTFT+VEL determined in the determination (step S13) of the voltage required by the organic EL element 121 and the driving transistor 125 and (ii) the voltage drop margin Vdrop determined in the determination (step S15) of the voltage margin corresponding to the potential difference  $\Delta V$ .

Lastly, the signal processing circuit 160 regulates the highside potential variable-voltage source 180 by setting the first reference voltage Vref1 as VTFT+VEL+Vdrop at the beginning of the next frame period (step S18). With this, in the next frame period, the high-side potential variable-voltage source 180 supplies Vout=VTFT+VEL+Vdrop to the organic EL display unit 110. It is to be noted that step S16 to step S18 correspond to the voltage regulating process according to the present disclosure.

In this manner, the display device 100 according to this embodiment includes: the high-side potential variable-voltage source 180 which outputs the high-side potential; the high-side potential difference detecting circuit 170 which measures, for the monitor pixel 111M in the organic EL display unit 110, (i) the high-side potential applied to the monitor pixel 111M and (ii) the high-side potential output voltage Vout of the high-side potential variable-voltage source 180; and the signal processing circuit 160 which regulates the high-side potential variable-voltage source 180 to set, to the predetermined potential (VTFT+VEL), the potential difference between the reference potential and the highside potential that is applied to the monitor pixel 111M that is measured by the high-side potential difference detecting circuit 170. Furthermore, the high-side potential difference detecting circuit 170, in addition, measures the high-side potential output voltage Vout of the high-side potential variable-voltage source 180, detects the potential difference between the measured high-side potential output voltage Vout and the high-side potential applied to the monitor pixel 111M. The signal processing circuit 160 regulates the high-side potential variable-voltage source 180 in accordance with the potential difference detected by the high-side potential difference detecting circuit 170.

With this, the display device 100 can reduce excess voltage and reduce power consumption by detecting the voltage drop caused by the horizontal first power source wire resistance R1h and the vertical first power source wire resistance R1v and giving feedback to the high-side potential variable-voltage source 180 regarding the degree of such voltage drop.

Furthermore, in the display device 100, the monitor pixel 111M is located near the center of the organic EL display unit 110, and thus the output voltage Vout of the high-side potential variable-voltage source 180 can be easily regulated even when the size of the organic EL display unit 110 is increased.

Furthermore, since heat generation by the organic EL element 121 is suppressed through the reduction of power consumption, the deterioration of the organic EL element 121 can be prevented.

Next, the display pattern transition in the case where the video data inputted up to the Nth frame changes from the N+1th frame onward, in the display device 100 described above, shall be described using FIG. 8 and FIG. 9.

Initially, the video data that is assumed to have been inputted in the Nth frame and the N+1th frame shall be described.

First, it is assumed that, up to the Nth frame, the video data corresponding to the central part of the organic EL display unit 110 is a peak gradation level (R:G:B=255:255:255) in which the central part of the organic EL display unit 110 is seen as being white. On the other hand, it is assumed that the video data corresponding to a part of the organic EL display unit 110 other than the central part is a gray gradation level (R:G:B=50:50:50) in which the part of the organic EL display unit 110 other than the central part is seen as being gray.

Furthermore, from the N+1th frame onward, it is assumed that the video data corresponding to the central part of the organic EL display unit 110 is the peak gradation level (R:G: B=255:255:255) as in the Nth frame. On the other hand, it is assumed that the video data corresponding to the part of the 25 organic EL display unit 110 other than the central part is a gray gradation level (R:G:8=150:150:150) that can be seen as a brighter gray than in the Nth frame.

Next, the operation of the display device 100 in the case where video data as described above is inputted in the Nth 30 frame and the N+1th frame shall be described.

The potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170, the output voltage Vout from the high-side potential variable-voltage source 180, and the pixel luminance of the monitor pixel 111M are 35 shown in FIG. 8. Furthermore, a blanking period is provided at the end of each frame period.

In time t=T10, the peak signal detecting circuit 150 detects the peak value of the video data of the Nth frame. The signal processing circuit 160 determines VTFT+VEL from the peak 40 value detected by the peak signal detecting circuit 150. Here, since the peak value of the video data of the Nth frame is R:G:B=255:255:255, the signal processing circuit 160 uses the required voltage conversion table and determines the required voltage VTFT+VEL for the N+1th frame to be, for 45 example, 12.2V.

Meanwhile, at this time, the high-side potential difference detecting circuit 170 detects the potential at the detecting point M1 via the monitor wire 190, and detects the potential difference  $\Delta V$  between the detected potential and the output 50 voltage Vout outputted from the high-side potential variablevoltage source 180. For example, in time t=T10, the high-side potential difference detecting circuit 170 detects  $\Delta V=1V$ . Subsequently, the signal processing circuit 160 uses the voltmargin Vdrop for the N+1th frame to be 1 V.

A time t=T10 to T11 is the blanking period of the Nth frame. In this period, an image which is the same as that in the time t=T10 is displayed in the organic EL display unit 110.

(a) in FIG. 9 schematically shows an image displayed on 60 the organic EL display unit 110 in the time t=T10 to T11. In this period, the image displayed on the organic EL display unit 110 corresponds to the image data of the Nth frame, and thus the central part is white and the part other than the central part is gray.

In time t=T11, the signal processing circuit 160 sets the voltage of the first reference voltage Vref1 as the sum VTFT+

22

VEL+Vdrop (for example, 13.2 V) of the determined required voltage VTFT+VEL and the voltage drop margin

Over a time t=T11 to T16, the image corresponding to the video data of the N+1th frame is gradually displayed on the organic EL display unit 110 ((b) to (f) in FIG. 9). At this time, the output voltage Vout from the high-side potential variablevoltage source 180 is, at all times, the VTFT+VEL+Vdrop set to the voltage of the first reference voltage Vref1 in time t=T11. However, the video data corresponding to the part of the organic EL display unit 110 other than the central part is a gray gradation level that can be seen as a gray that is brighter than that in the Nth frame. Therefore, the amount of current supplied by the high-side potential variable-voltage source 180 to the organic EL display unit 110 gradually increases over time t=T11 to T16, and the voltage drop in the first power source wire 112 gradually increases following this increase in the amount of current. With this, there is a shortage of power source voltage for the pixels 111 in the central part of the 20 organic EL display unit 110, which are the pixels 111 in a brightly displayed region. Stated differently, luminance drops below the image corresponding to the video data R:G:B=255: 255:255 of the N+1th frame. Specifically, over the time t=T11 to T16, the pixel luminance of the pixels 111 at the central part of the organic EL display unit 110 gradually drops.

Next, in time t=T16, the peak signal detecting circuit 150 detects the peak value of the video data of the N+1th frame. Here, since the detected peak value of the video data of the N+1th frame is R:G:B=255:255:255, the signal processing circuit 160 determines the required voltage VTFT+VEL for the N+2th frame to be, for example, 12.2 V.

Meanwhile, at this time, the high-side potential difference detecting circuit 170 detects the potential at the detecting point M1 via the monitor wire 190, and detects the potential difference  $\Delta V$  between (i) the potential difference between the detected potential at the detecting point M1 and the reference potential and (ii) and the output voltage Vout outputted from the high-side potential variable-voltage source 180. For example, in time t=T16, the high-side potential difference detecting circuit 170 detects  $\Delta V=3$  V. Subsequently, the signal processing circuit 160 uses the voltage margin conversion table and determines the voltage drop margin Vdrop for the N+1th frame to be 3 V.

Next, in time t=T17, the signal processing circuit 160 sets the voltage of the first reference voltage Vref1 to the sum VTFT+VEL+Vdrop (for example, 15.2 V) of the determined required voltage VTFT+VEL and the voltage drop margin Vdrop. Therefore, from time t=T17 onward, the potential at the detecting point M1 becomes VTFT+VEL which is the predetermined potential.

In this manner, in the display device 100, although luminance temporarily drops in the N+1th frame, this is a very short period and thus has practically no impact on the user.

It is to be noted that although, in this embodiment, the age margin conversion table and determines the voltage drop 55 high-side potential applied to the pixel 111M is monitored and the high-side potential outputted from the high-side potential variable-voltage source 180 is regulated to set the potential difference between the monitored high-side potential and the reference potential to a predetermined voltage, the low-side potential applied to the pixel 111M may be monitored and the low-side potential outputted from a low-side potential variable-voltage source may be regulated to set the potential difference between the monitored low-side potential and the reference potential to a predetermined voltage. In this case, a low-side potential difference detecting circuit is provided in place of the high-side potential difference detecting circuit 170 shown in FIG. 1, and a low-side potential

variable-voltage source is provided in place of the high-side potential variable-voltage source 180.

A peak signal detecting circuit detects the peak value of video data inputted to the display device 100, and outputs a peak signal indicating the detected peak value to the signal 5 processing circuit. Specifically, the functions of such peak signal detecting circuit are the same as those of the peak signal detecting circuit 150.

The signal processing circuit regulates the low-side potential variable-voltage source in accordance with the peak signal outputted by the peak signal detecting circuit and the potential difference  $\Delta V$  detected by the low-side potential difference detecting circuit, to set the potential of the monitor pixel 111M to a predetermined potential.

The low-side potential difference detecting circuit mea- 15 sures, for the monitor pixel 111M, a low-side potential applied to the monitor pixel 111M. In addition, the low-side potential difference detecting circuit measures the output potential of the low-side potential variable-voltage source, and measures the potential difference  $\Delta V$  between (i) the  $^{20}$ potential difference between the measured low-side potential applied to the monitor pixel 111M and the reference potential and (ii) the output potential of the low-side variable-voltage source. Subsequently, the low-side potential difference detecting circuit outputs the measured potential difference 25  $\Delta V$  to the signal processing circuit.

The low-side potential variable-voltage source outputs the low-side potential to the organic EL display unit 110. The low-side potential variable-voltage source outputs an output voltage Vout for setting the potential difference between the 30 low-side potential of the monitor pixel 111M and the reference potential to a predetermined voltage, in accordance with the reference voltage Vref outputted by the signal processing circuit. It is sufficient that reference potential be a potential serving as a reference in the display device 100.

With this, the display device in the present disclosure can reduce excess voltage and reduce power consumption by detecting the voltage drop caused by the horizontal second power source wire resistance R2h and the vertical second power source wire resistance R2v and giving feedback to the 40 low-side potential variable-voltage source regarding the degree of such voltage drop.

Furthermore, since heat generation by the organic EL element is suppressed through the reduction of power consumption, the deterioration of the organic EL element can be pre- 45 vented.

#### **Embodiment 3**

A display device according to this embodiment is almost 50 the same as the display device 100 according to Embodiment 2, but is different in not including the high-side potential difference detecting circuit 170 and in that the potential at the detecting point M1 is inputted to the high-side potential variis different in setting the voltage to be outputted to the highside potential variable-voltage source to the required voltage VTFT+VEL. With this, in the display device according to this embodiment, the output voltage Vout of the high-side potential variable-voltage source can be regulated in real-time in 60 accordance with the voltage drop amount, and thus, compared with Embodiment 2, the temporary drop in pixel luminance can be prevented.

FIG. 14 is a block diagram showing an outline configuration of the display device according to Embodiment 3.

A display device 200 according to the this embodiment shown in the figure is different compared to the display device 24

100 according to Embodiment 2 shown in FIG. 10 in not including the high-side potential difference detecting circuit 170, and including a monitor wire 290 in place of the monitor wire 190, a signal processing circuit 260 in place of the signal processing circuit 160, and a high-side potential variablevoltage source 280 in place of the high-side potential variable-voltage source 180.

The signal processing circuit 260 determines a second reference voltage Vref2 to be outputted to the high-side potential variable-voltage source 280, from the peak signal outputted by a peak signal detecting circuit 150. Specifically, the signal processing circuit 260 uses the required voltage conversion table and determines the sum VTFT+VEL of the voltage VEL required by an organic EL element 121 and a voltage VTFT required by the driving transistor 125. Subsequently, the signal processing circuit 260 sets the determined VTFT+VEL as the voltage of the second reference voltage

In such manner, the second reference voltage Vref2 that is outputted to the high-side potential variable-voltage source 280 by the signal processing circuit 260 of the display device 200 according to this embodiment is different from the first reference voltage Vref1 that is outputted to the high-side potential variable-voltage source 180 by the signal processing circuit 160 of the display device 100 according to Embodiment 2, and is a voltage determined in accordance with the video data only. Specifically, the second reference voltage Vref2 is not dependent on the potential difference  $\Delta V$ between the potential of the output voltage Vout of the highside potential variable-voltage source 280 and the potential at the detecting point M1.

The high-side potential variable-voltage source 280 measures the high-side potential applied to the monitor pixel 111M, via the monitor wire 290. Specifically, the high-side potential variable-voltage source 280 measures the potential at the detecting point M1. Subsequently, the high-side potential variable-voltage source 280 regulates the output voltage Vout in accordance with (i) the potential difference between the measured potential at the detecting point M1 and the reference potential and (ii) the second reference voltage Vref2 outputted by the signal processing circuit 260.

The monitor wire 290 has one end connected to the detecting point M1 and the other end connected to the high-side potential variable-voltage source 280, and transmits the potential at the detecting point M1 to the high-side potential variable-voltage source 280.

FIG. 15 is a block diagram showing an example of a specific configuration of the high-side potential variable-voltage source 280 according to Embodiment 3. It is to be noted that the organic EL display unit 110 and the signal processing circuit 260 which are connected to the high-side potential variable-voltage source are also shown in the figure.

The high-side potential variable-voltage source 280 shown able-voltage source. Furthermore, a signal processing circuit 55 in the figure has almost the same configuration as the highside potential variable-voltage source 180 shown in FIG. 11, but is different in including, in place of the comparison circuit 181, a comparison circuit 281 which compares the potential at the detecting point M1 and the second reference voltage Vref2.

> Here, assuming that the output potential of the high-side potential variable-voltage source 280 is Vout, and the voltage drop amount from the output terminal **184** of the high-side potential variable-voltage source 280 to the detecting point M1 is  $\Delta V$ , the potential at the detecting point M1 is Vout- $\Delta V$ . Specifically, in this embodiment, the comparison circuit 281 compares Vref2 and Vout-ΔV. As described above, since

Vref2=VTFT+VEL, it can be said that the comparison circuit **281** is comparing VTFT+VEL and Vout- $\Delta V$ .

On the other hand, in Embodiment 2, the comparison circuit **181** compares Vref1 and Vout. As described above, since Vref1=VTFT+VEL+ $\Delta$ V, it can be said that, in Embodiment 52, the comparison circuit **181** is comparing VTFT+VEL+ $\Delta$ V and Vout.

Therefore, although the comparison circuit **281** has different comparison subjects as the comparison circuit **181**, the comparison result is the same. Specifically, when the voltage 10 drop amount from the output terminal **184** of the high-side potential variable-voltage source to the detecting point **M1** is the same between Embodiment 2 and Embodiment 3, the voltage outputted by the comparison circuit **181** to the PWM circuit and the voltage outputted by the comparison circuit 15 **281** to the PWM circuit are the same. As a result, the output voltage Vout of the high-side potential variable-voltage source **180** and the output voltage Vout of the high-side potential variable-voltage source **280** become the same. Furthermore, the potential difference  $\Delta V$  and the output voltage Vout also have an increasing function relationship in Embodiment 3.

Compared to the display device 100 according to Embodiment 2, the display device 200 configured in the above manner can regulate the output voltage Vout in accordance with 25 the potential difference  $\Delta V$  between the output terminal 184 and the detecting point M1 in real-time. This is because, in the display device 100 according to Embodiment 2, the signal processing circuit 160 changes the first reference voltage Vref1 for a frame only at the beginning of each frame period. 30 In contrast, in the display device 200 according to this embodiment, Vout can be regulated independently of the control by the signal processing circuit 260, by inputting the voltage that is dependent on the  $\Delta V$ , that is, Vout- $\Delta V$  directly to the comparison circuit 281 of the high-side potential variable-voltage source 280 without passing through the signal processing circuit 260.

Next, the operation of the display device **200** configured in the above manner, in the case where the video data inputted up to the Nth frame changes from the N+1th frame onward, as in Embodiment 2, shall be described. It is to be noted that, as in Embodiment 2, it is assumed that, up to the Nth frame, the inputted video data is R:G:B=255:255:255 for the central part of the organic EL display unit **110** and is R:G:B=50:50:50 for the part other than the central part, and, from the N+1th frame 45 onward, the inputted video data is R:G:B=255:255:255 for the central part of the organic EL display unit **110** and is R:G:B=150:150:150 for the part other than the central part.

FIG. 16 is a timing chart showing the operation of the display device 200 from the Nth frame to the N+2th frame. 50

In time t=T20, the peak signal detecting circuit 150 detects the peak value of the video data of the Nth frame. The signal processing circuit 260 determines VTFT+VEL from the peak value detected by the peak signal detecting circuit 150. Here, since the peak value of the video data of the Nth frame is 55 R:G:B=255:255; the signal processing circuit 260 uses the required voltage conversion table and determines the required voltage VTFT+VEL for the N+1th frame to be, for example, 12.2V.

Meanwhile, the output detecting unit 185 constantly 60 detects the potential at the detecting point M1, via the monitor wire 290.

Next, in time t=T21, the signal processing circuit 260 sets the voltage of the second reference voltage Vref2 to the determined required voltage VTFT+VEL (for example, 12.2V).

Over a time t=T21 to T22, the image corresponding to the video data of the N+1th frame is gradually displayed on the

26

organic EL display unit 110. At this time, the amount of current supplied by the high-side potential variable-voltage source 280 to the organic EL display unit 110 gradually increases, as described in Embodiment 1. Therefore, following the increase in the amount of current, the voltage drop in the first power source wire 112 gradually increases. Specifically, the potential at the detecting point M1 gradually drops. Stated differently, the potential difference  $\Delta V$  between the potential of the output voltage Vout and the potential at the detecting point M1 gradually increases.

Here, since the error amplifier 186 outputs, in real-time, a voltage that is in accordance with the potential difference between VTFT+VEL and Vout- $\Delta V$ , the error amplifier 186 outputs a voltage that causes Vout to rise in accordance with the increase in the potential difference  $\Delta V$ .

Therefore, with the high-side potential variable-voltage source **280**, Vout rises in real-time in accordance with the increase in the potential difference  $\Delta V$ .

This resolves the shortage of power source voltage for the pixels 111 in the central part of the organic EL display unit 110 which are the pixels 111 in the brightly displayed region. In other words, the drop in pixel luminance is resolved.

As described above, in the display device 200 according to the this embodiment, the signal processing circuit 260, and the error amplifier 186, the PWM circuit 182, and the drive circuit 183 of the high-side potential variable-voltage source 280, detect the potential difference between (i) the potential difference between the high-side potential of the monitor pixel 111M measured by the output detecting unit 185 and the reference potential and (ii) the predetermined potential, and regulates the switching element SW in accordance with the detected potential difference. Accordingly, compared with the display device 100 according to Embodiment 2, the display device 200 according to this embodiment is able to regulate the output voltage Vout of the high-side potential variable-voltage source 280 in real-time in accordance with the voltage drop amount, and thus compared to Embodiment 2, the temporary drop in pixel luminance can be prevented.

It is to be noted that, in this embodiment, the organic EL display unit 110 is the display unit; the output detecting unit 185 is the voltage detecting unit; the signal processing circuit 260, and the error amplifier 186, PWM circuit 182, and drive circuit 183 of the high-side potential variable-voltage source 280 which are surrounded by the dashed-and-single-dotted line in FIG. 15 are the voltage regulating unit; and the switching element SW, the diode D, the inductor L, and the capacitor C which are surrounded by the dashed-and-double-dotted line in FIG. 15 are the power supplying unit.

It is to be noted that, in the same manner as in Embodiment 2, in this embodiment, the low-side potential applied to the pixel 111M may be monitored and the low-side potential outputted from a low-side potential variable-voltage source may be regulated to set the potential difference between the monitored low-side potential and the reference potential to a predetermined voltage. In this case, a low-side potential variable-voltage source is provided in place of the high-side potential variable-voltage source 180 shown in FIG. 14, and a low-side potential variable-voltage source is provided in place of the high-side potential variable-voltage source source 180.

## Embodiment 4

Embodiments 2 and 3 describe a display device which monitors the high-side or low-side potential of a single pixel to regulate, to a predetermined potential difference, the potential difference between the high-side potential and a reference potential or the potential difference between the low-side

potential and the reference potential. In contrast, this embodiment describes a display device which monitors the high-side potential of a single pixel and the low-side potential of a different pixel to regulate the potential difference between the high-side potential and a reference potential A to a predetermined potential difference, and to regulate the potential difference between the low-side potential and a reference potential B to a predetermined potential difference.

Hereinafter, Embodiment 4 shall be specifically described with reference to the Drawings.

FIG. 17 is a block diagram showing an outline configuration of a display device according to Embodiment 4.

A display device 300 shown in the figure includes an organic EL display unit 310, the data line driving circuit 120, the write scan driving circuit 130, the control circuit 140, the signal processing circuit 165, a high-side potential difference detecting circuit 170A, a low-side potential difference detecting circuit 170B, a high-side potential voltage margin setting unit 175A, a low-side potential voltage margin setting unit 175B, a high-side potential variable-voltage source 180A, a 20 low-side potential variable-voltage source 180B, and monitor wires 190A and 190B.

Compared to the display device **50** according to Embodiment 1, the display device **300** according to this embodiment is different in including two potential difference detecting 25 circuits, two monitor wires, and two variable-voltage sources, for the high-side potential and the low-side potential, respectively. Description of points identical to those in Embodiment 1 shall not be repeated, and only the points of difference shall be described hereafter.

FIG. 18 is a perspective view schematically showing a configuration of the organic EL display unit 310 according to Embodiment 4. It is to be noted that the lower portion of the figure is the display screen side. As shown in the figure, the organic EL display unit 310 includes the pixels 111, the first power source wire 112, and the second power source wire 113. At least one predetermined pixel out of the pixels 111 is connected to the monitor wire 190A at a high-side potential detecting point  $M_A$ . Furthermore, at least one predetermined pixel out of the pixels 111 is connected to the monitor wire 190B at a low-side potential detecting point  $M_B$ . Hereinafter, the pixel 111 that is directly connected to the monitor wire 190A shall be denoted as monitor pixel  $111M_A$ , and the pixel 111 that is directly connected to the monitor wire 190B shall be denoted as monitor pixel  $111M_B$ .

The first power source wire 112 is arranged in a net-like manner to correspond to the pixels 111 that are arranged in a matrix, and is electrically connected to the high-side potential variable-voltage source 180A disposed at the periphery of the organic EL display unit 310. Through the outputting of a 50 high-side power source potential from the high-side potential variable-voltage source 180A, a potential corresponding to the high-side power source potential outputted by the highside potential variable-voltage source 180A is applied to the first power source wire 112. On the other hand, the second 55 power source wire 113 is formed in the form of a continuous film on the organic EL display unit 310, and is connected to the low-side potential variable-voltage source 180B disposed at the periphery of the organic EL display unit **310**. Through the outputting of a low-side power source potential from the 60 low-side potential variable-voltage source 180A, a potential corresponding to the low-side power source potential outputted by the low-side potential variable-voltage source **180**A is applied to the second power source wire 113.

The optimal position of the monitor pixels  $111M_A$  and 65  $111M_B$  is determined depending on the wiring method of the first power source wire 112 and the second power source wire

28

113, the respective values of the first power source wire resistances R1h and R1 $\nu$ , and the respective values of the second power source wire resistances R2h and R2 $\nu$ . In this embodiment, the high-side potential detecting point M $_A$  and the low-side potential detecting point M $_B$  are disposed in different pixels. This allows for optimization of detecting points. For example, by disposing the pixel  $111M_A$  in a luminescence producing region in which there is a tendency for a large high-side potential voltage drop, and disposing the pixel  $111M_B$  in a luminescence producing region in which there is a tendency for a large low-side potential voltage drop (rise), detecting points need not be provided in unnecessary locations and thus the total number of detecting points can be reduced.

Since a cathode electrode of an organic EL element 121 which makes up part of a common electrode included in the second power source wire 113 uses a transparent electrode (for example, ITO) having high sheet resistance, there are cases where the voltage rise amount for the second power source wire 113 is larger than the voltage drop amount for the first power source wire 112. Therefore, by regulating in accordance with the low-side potential applied to the monitor pixel, the output potential of the power supplying unit can be regulated more appropriately, and power consumption can be further reduced.

FIG. 19A and FIG. 19B are circuit diagrams showing an example of a specific configuration of a pixel 111. Specifically, FIG. 19A is a diagram of the circuit configuration of the pixel 111M<sub>4</sub> connected to the high-side potential monitor wire 190A, and FIG. 19B is a diagram of the circuit configuration of the pixel  $111M_B$  connected to the low-side potential monitor wire 190B. In the pixel 111M<sub>4</sub>, the monitor wire 190A is connected to the other of the source electrode and the drain electrode of the driving element, and, in the pixel 111M<sub>B</sub>, the monitor wire 190B is connected to the second electrode of the luminescence element. Specifically, each of the pixels 111, 111M<sub>4</sub>, and 111M<sub>B</sub> includes an organic EL element 121, a data line 122, a scanning line 123, a switch transistor 124, a driving transistor 125, and a holding capacitor 126. At least one pixel 111M<sub>4</sub> is disposed in the organic EL display unit 310, and at least one pixel  $111M_B$  is likewise disposed in the organic EL display unit 310.

The functions of the respective constituent elements shown in FIG. 17 shall be described below with reference to FIG. 19A and FIG. 19B.

The high-side potential difference detecting circuit 170A, which is the voltage detecting unit in this embodiment, measures, for the monitor pixel 111M<sub>4</sub>, the high-side potential to be applied to the monitor pixel 111M<sub>4</sub>. Specifically, the highside potential difference detecting circuit 170A measures, via the monitor wire 190A, the high-side potential to be applied to the monitor pixel  $111M_A$ . In addition, the high-side potential difference detecting circuit 170A measures the output potential of the high-side potential variable-voltage source **180**A, and measures the potential difference  $\Delta VH$  between (i) the potential difference between the measured high-side potential to be applied to the monitor pixel 111M<sub>4</sub> and the reference potential A and (ii) the output potential of the highside potential variable-voltage source 180A. Subsequently, the high-side potential difference detecting circuit 170A outputs the measured potential difference ΔVH to the high-side potential voltage margin setting unit 175A.

The low-side potential difference detecting circuit 170B, which is the voltage detecting unit in this embodiment, measures, for the monitor pixel  $111\mathrm{M}_B$ , the low-side potential to be applied to the monitor pixel  $111\mathrm{M}_B$ . Specifically, the low-side potential difference detecting circuit 170B measures, via

the monitor wire  $190\mathrm{B}$ , the low-side potential to be applied to the monitor pixel  $111\mathrm{M}_B$ . In addition, the low-side potential difference detecting circuit  $170\mathrm{B}$  measures the output potential of the low-side potential variable-voltage source  $180\mathrm{B}$ , and measures the potential difference  $\Delta\mathrm{VL}$  between (i) the potential difference between the measured low-side potential to be applied to the monitor pixel  $111\mathrm{M}_B$  and the reference potential B and (ii) the output potential of the low-side variable-voltage source  $180\mathrm{B}$ . Subsequently, the low-side potential difference detecting circuit  $170\mathrm{B}$  outputs the measured potential difference  $\Delta\mathrm{VL}$  to the low-side potential voltage margin setting unit  $175\mathrm{B}$ .

The high-side potential voltage margin setting unit 175A, which is the high-side potential voltage regulating unit in this embodiment, regulates, based on a voltage (VEL+VTFT) at a 15 peak gradation level and the potential difference  $\Delta VH$ detected by the high-side potential difference detecting circuit 170A, the high-side potential variable-voltage source 180A to set the potential difference between the potential of the monitor pixel 111M<sub>4</sub> and the reference potential A to a 20 predetermined potential. Specifically, the high-side potential voltage margin setting unit 175A calculates a voltage drop margin VHdrop based on the potential difference detected by the high-side potential difference detecting circuit 170A. Subsequently, the high-side potential voltage margin setting 25 unit 175A sums up the voltage (VEL+VTFT) at the peak gradation level and the voltage drop margin VHdrop, and outputs a higher voltage than the reference potential A of the summation result VEL+VTFT+VHdrop, as a first high-side potential reference voltage VHref1, to the high-side potential 30 variable-voltage source 180A.

The low-side potential voltage margin setting unit 175B, which is the low-side potential voltage regulating unit in this embodiment, regulates, based on a voltage (VEL+VTFT) at a peak gradation level and the potential difference  $\Delta VL$  35 detected by the low-side potential difference detecting circuit 170B, the low-side potential variable-voltage source 180B to set the potential difference between the potential of the monitor pixel 111M<sub>B</sub> and the reference potential B to a predetermined potential. Specifically, the low-side potential voltage 40 margin setting unit 175B calculates a voltage drop margin VLdrop based on the potential difference detected by the low-side potential difference detecting circuit 170B. Subsequently, the low-side potential voltage margin setting unit 175B sums up the voltage (VEL+VTFT) at the peak gradation 45 level and the voltage drop margin VLdrop, and outputs a lower voltage than the reference potential B of the summation result VEL+VTFT+VLdrop, as a first low-side potential reference voltage VLref1, to the low-side potential variablevoltage source 180B.

The high-side potential variable-voltage source  $180\,\mathrm{A}$ , which is the power supplying unit in this embodiment, outputs the high-side potential to the organic EL display unit 310. The high-side potential variable-voltage source  $180\,\mathrm{A}$  outputs an output voltage VHout for setting the potential difference 55 between the high-side potential of the monitor pixel  $111\,\mathrm{IM}_A$  and the reference potential A to the predetermined voltage (VEL+VTFT-reference potential A), according to the first high-side potential reference voltage VHref1 outputted by the high-side potential voltage margin setting unit  $175\,\mathrm{A}$ . It is sufficient that reference potential A be a potential serving as a reference in the display device 300.

The low-side potential variable-voltage source **180**B, which is the power supplying unit in this embodiment, outputs the low-side potential to the organic EL display unit **310**. 65 The low-side potential variable-voltage source **180**B outputs an output voltage VLout for setting the potential difference

30

between the low-side potential of the monitor pixel  $111M_B$  and the reference potential B to the predetermined voltage (reference potential B-VEL+VTFT), according to the first low-side potential reference voltage VLref1 outputted by the low-side potential voltage margin setting unit 175B.

The monitor wire 190A is a high-side potential detecting line which has one end connected to the monitor pixel  $111M_A$  and the other end connected to the high-side potential difference detecting circuit 170A, and transmits the high-side potential applied to the monitor pixel  $111M_A$  to the high-side potential difference detecting circuit 170A.

The monitor wire 190B is a low-side potential detecting line which has one end connected to the monitor pixel  $111M_B$  and the other end connected to the low-side potential difference detecting circuit 170B, and transmits the low-side potential applied to the monitor pixel  $111M_B$  to the low-side potential difference detecting circuit 170B.

Furthermore, the configuration of the high-side potential variable-voltage source 180A and the low-side potential variable-voltage source 180B according this embodiment is the same as the configuration of the high-side potential variable-voltage source 180 according to Embodiment 1. In the case where the output voltage VLout of the low-side potential variable-voltage source 180B is negative, the circuit of the low-side potential variable-voltage source 180B is configured by changing the arrangement of the switching element SW, the diode D, the inductor L, and the capacitor C in FIG. 4.

Furthermore, with regard to the operation of the display device 300 according to this embodiment, the operation from step S14 to step S18 in FIG. 5 describing the operational flow for the display device 50 in Embodiment 1 is executed in parallel for the high-side potential and the low-side potential.

According to this embodiment, the display device 300 can reduce excess voltage and reduce power consumption by detecting the voltage drop caused by the first power source wire resistance R1h and the first power source wire resistance  $R1\nu$  in the positive power source line-side and the voltage rise caused by the second power source wire resistance R2h and the second power source wire resistance  $R2\nu$  in the negative power source line-side, and giving feedback to the high-side potential variable-voltage source 180A and the low-side potential variable-voltage source 180A regarding the degree of such voltage drop and voltage rise, respectively.

Furthermore, since heat generation by the organic EL element **121** is suppressed through the reduction of power consumption, the deterioration of the organic EL element **121** can be prevented.

In addition, compared to the display device in Embodiment 1, in the display device 300 according to this embodiment, it is possible to set a voltage margin that takes into consideration a voltage rise that is proportionate to the wire resistance of the low-side potential power source line, and thus power consumption can be more effectively reduced in a display mode in which the voltage distribution of the low-side potential power source line is intense.

It is to be noted that in Embodiments 1 to 4, the output voltage from the variable-voltage source is regulated based on the potential difference between the voltage applied to the pixels and the voltage outputted by the variable-voltage source. In this case, the current path from the variable-voltage source to the pixels includes a wiring path outside the display region and a wiring path inside the display region in which the pixels are disposed. Specifically, in Embodiments 1 to 4, the output voltage from the variable-voltage source is regulated in accordance with the voltage drop amount both inside the display region and outside the display region, by detecting the potential difference between the voltage applied to the pixels

and the voltage outputted from the variable-voltage source. In contrast, the output voltage from the variable-voltage source can be regulated in accordance with the voltage drop (rise) amount inside the display region only, by detecting the potential difference between the voltage applied to the pixels and the voltage in the wiring path outside the display region. This shall be described below using FIG. 20A and FIG. 20B.

FIG. 20A is a diagram showing an outline configuration of a display panel included in a display device disclosed herein. Furthermore, FIG. 20B is perspective diagram schematically showing the vicinity of the periphery of the display panel included in a display device disclosed herein. In FIG. 20A, drivers such as write scan driving circuits and data line driving circuits, high-side potential power source lines, low-side 15 potential power source lines, and flexible pads, which are interfaces for electrical connection with outside devices, are disposed in the periphery of a display panel in which pixels 111 are arranged in a matrix. Each of the high-side potential variable-voltage source and the low-side potential variable- 20 voltage source is connected to the display panel via (i) a high-side potential power source line and flexible pads or (ii) a low-side potential power source line and flexible pads. As shown in FIG. 20B, resistance components are also present outside the display region, and such resistance components 25 are due to the aforementioned flexible pads, high-side potential power source lines and low-side potential power source lines. Although in Embodiments 1 to 4 described earlier, for example, the difference between the potential at the detecting point  $M_A$  and the potential of an output point Z of the highside potential variable-voltage source is detected, the potential difference between the potential at the detecting point M<sub>4</sub> and the potential at a connection point Y of the display panel and a high-side potential power source line may be detected for the purpose of regulating the output voltage from the 35 variable-voltage source that is in accordance with the voltage drop amount only inside the display region. With this, the output voltage of the variable-voltage source can regulated in accordance with the voltage drop amount within the display region only. Furthermore, for the low-side potential, the 40 potential difference between the potential at the detecting point  $M_B$  and the potential at a connection point of the display panel and a low-side potential power source line may be detected.

It is to be noted that although this embodiment describes a 45 display device which monitors the high-side potential of one pixel and the low-side potential of a different pixel to thereby (i) regulate the potential difference between the high-side potential and the reference potential A to a predetermined potential difference and (ii) regulate the potential difference 50 between the low-side potential and the reference potential B to a predetermined potential difference, the pixel from which the high-side potential is detected and the pixel from which the low-side potential is detected may be the same pixel. Even in such a case, the high-side potential variable-voltage source 55 180A regulates the potential difference between the high-side potential and the reference potential A to a predetermined potential difference, and the low-side potential variable-voltage source 180B regulates the potential difference between the low-side potential and the reference potential B to a pre- 60 determined potential difference.

#### Embodiment 5

This embodiment describes a display device that monitors 65 the high-side potentials of plural pixels to thereby regulate, to a predetermined potential difference, the potential difference

32

between a high-side potential specified from among the monitored high-side potentials and the reference potential.

Hereinafter, Embodiment 5 shall be specifically described with reference to the Drawings.

FIG. 21 is a block diagram showing an outline configuration of a display device according to Embodiment 5.

A display device 400 shown in the figure includes an organic EL display unit 410, the data line driving circuit 120, the write scan driving circuit 130, the control circuit 140, the peak signal detecting circuit 150, the signal processing circuit 160, the high-side potential difference detecting circuit 170, the high-side potential variable-voltage source 180, monitor wires 191, 192, and 193, and a potential comparison circuit 470.

Compared to the display device **100** according to Embodiment 2, the display device **400** according to this embodiment is different in including plural monitor wires and the potential comparison circuit **470**. Description of points identical to those in Embodiment 2 shall not be repeated, and only the points of difference shall be described hereafter.

The organic EL display unit **410** is nearly the same as the organic EL display unit **110** but is different compared to the organic EL display unit **110** in the placement of the monitor wires **191** to **193** which are provided, on a one-to-one correspondence with detecting points **M1** to **M3**, for measuring the potential at the corresponding detecting point.

The optimal position of the monitor pixels 111M to 111M3 is determined depending on the wiring method of the first power source wire 112, and the respective values of the first power source wire resistances R1h and R1v.

Each of the monitor wires 191 to 193 is connected to the corresponding one of the detecting points M1 to M3 and to the potential comparison circuit 470, and transmits the potential at the corresponding one of the detecting points M1 to M3 to the potential comparison circuit 470. With this, the potential comparison circuit 470 can measure the potentials at the detecting points M1 to M3 via the monitor wires 191 to 193.

The potential comparison circuit 470 measures the potentials at the detecting points M1 to M3 via the corresponding ones of the monitor wires 191 to 193. Stated differently, the potential comparison circuit 470 measures the high-side potential applied to the monitor pixels 111M1 to 111M3. In addition, the potential comparison circuit 470 selects the lowest potential among the measured potentials at the detecting points M1 to M3, and outputs the selected potential to the high-side potential difference detecting circuit 170.

The signal processing unit 160 regulates the high-side potential variable-voltage source 180 based on the potential difference between the potential selected by the potential comparison circuit 470 and the reference potential. As a result, the high-side potential variable-voltage source 180 provides, to the organic EL display unit 410, an output voltage Vout with which dropping of luminance does not occur in any of the monitor pixels 111M1 to 111M3.

As described above, in the display device 400 according to this embodiment, the potential comparison circuit 470 measures the high-side potential applied to each of the pixels 111 inside the organic EL display unit 410, and selects the lowest potential among the measured high-side potentials. In addition, the high-side potential difference detecting circuit 170 detects the potential difference  $\Delta V$  between (i) the potential difference between the lowest potential selected by the potential comparison circuit 470 and the reference potential and (ii) the potential of the output voltage Vout of the high-side potential variable-voltage source 180. Then, the signal processing

circuit 160 regulates the high-side potential variable-voltage source 180 in accordance with the detected potential difference  $\Delta V$ .

With this, the output voltage Vout of the high-side potential variable-voltage source 180 can be more appropriately regulated. Therefore, power consumption can be effectively reduced even when the size of the organic EL display unit is increased

It is to be noted that, in the display device **400** according to this embodiment: the high-side potential variable-voltage source **180** is the power supplying unit; the organic EL display unit **410** is the display unit; one part of the potential comparison circuit **470** is the voltage detecting unit; and the other part of the potential comparison circuit **470**, the high-side potential difference detecting circuit **170**, and the signal processing circuit **160** are the voltage regulating unit.

Furthermore, although the potential comparison circuit 470 and the high-side potential difference detecting circuit 170 are provided separately in the display device 400, a 20 potential comparison circuit which compares the potential of the output voltage Vout of the variable-voltage source 180 and the potential at each of the detecting points M1 to M3 may be provided in place of the potential comparison circuit 470 and the high-side potential difference detecting circuit 170.

Next, the advantageous effects produced by the display device 400 according to this embodiment shall be described.

FIG. 22 is a diagram showing potential distributions and a detecting point arrangement for the display device according to Embodiment 5. The diagram on the left side of FIG. 22 30 shows the potential distributions when 15 V is applied as the high-side potential power source output and 0 V, which is a grounding potential, is applied as the low-side potential power source output. Since a 1:10 ratio is assumed between the first power source wire resistance R1h and the first power 35 source resistance R1V, the high-side potential distribution shows a severe potential change in the vertical direction of the display panel. In contrast, since a 10:1 ratio is assumed between the second power source wire resistance R2h and the second power source resistance R2V, the low-side potential 40 distribution shows a small potential change over the entire display panel. In other words, the low-side potential distribution has a tendency to be approximately uniform within the display screen.

When such tendencies are present, it is possible, for 45 example, to monitor only the potential distribution of the high-side potential which has an extremely-varying distribution, and regulate the voltage drop (rise) amount of the low-side potential based on the potential of the high-side potential. Stated in terms of the example in FIG. 22, with respect to 50 maximum voltage drop amount detected from the potential distribution of the high-side potential which is 3 V (15 V–12 V), the voltage drop (rise) amount of the low-side potential is considered at all times to be half (1.5 V) of such detected drop amount (3 V).

In the display panel having the characteristics shown in FIG. 22, a significant error does not occur even when the above-described voltage drop (rise) amount of the low-side potential is not measured, and thus, as a result, there is the advantage of obtaining a power conservation effect while 60 reducing the detecting points for the low-side potential. Specifically, even without measuring the high-side potential and the low-side potential at each of the set pixels 111M1 to 111M3, it is sufficient to measure only the high-side potential at each of the pixels 111M1 to 111M3, and thus the detecting 65 points are reduced from six points to three points. With this, the design of the interior of the display panel which requires

34

consideration to the arrangement of monitor wires is simplified and picture quality deterioration due to monitor wire addition can be avoided.

In addition, since monitor wires for the low-side potential are eliminated, in a panel format in which light is emitted from the side at which the low-side potential is detected, there is the advantage that line defects originating from the monitor lines are not readily visible.

It is to be noted that although the three detecting points M1 to M3 are illustrated in the figure, having plural detecting points is sufficient, and it is sufficient to determine the optimal positioning and number of points based on the method of wiring of the power source wires and the value of the wire resistance.

Although the display device according to the present disclosure has been described thus far based on the embodiments, the display device according to the present disclosure is not limited to the above-described embodiments. Modifications that can be obtained by executing various modifications to Embodiments 1 to 5 that are conceivable to a person of ordinary skill in the art without departing from the essence of the present disclosure, and various devices internally equipped with the display device according to the present disclosure are included in the present disclosure.

For example, the drop in the pixel luminance of the pixel to which the monitor wire inside the organic EL display unit is provided may be compensated.

FIG. 23 is a graph showing the pixel luminance of a normal pixel and the pixel luminance of a pixel having the monitor wire, which correspond to the gradation levels of video data. It is to be noted that a normal pixel refers to a pixel among the pixels of the organic EL display unit, other than the pixel provided with a monitor wire.

As is clear from the figure, when the gradation levels of the video data are the same, the luminance of the pixel having the monitor wire drops more than the luminance of the normal pixel. This is because, with the provision of a monitor wire, the capacitance value of the holding capacitor 126 of the pixel decreases. Therefore, even when video data which causes luminance to be produced with the same luminance evenly throughout the entirety of the organic EL display unit is inputted, the image to be displayed on the organic EL display unit is an image in which the luminance of the pixels having a monitor wire is lower than the luminance of the other pixels. In other words, line defects occur. FIG. 24 is a diagram schematically showing an image in which line defects occur.

In order to prevent line defects, the display device may correct the signal voltage applied to the organic EL display unit from the data line driving circuit 120. Specifically, since the positions of the pixels having a monitor wire are known at the time of designing, it is sufficient to pre-set the signal voltage to be provided to the pixels in such locations to be higher by the amount of drop in luminance. With this, it is possible to prevent line defects caused by the provision of monitor wires.

Furthermore, although the signal processing circuit has the required voltage conversion table indicating the required voltage VTFT+VEL corresponding to the gradation levels of each color, the signal processing circuit may have, in place of the required voltage conversion table, the current-voltage characteristics of the driving transistor 125 and the current-voltage characteristics of the organic EL element 121, and determine VTFT+VEL by using these two current-voltage characteristics.

FIG. **25** is a graph showing together current-voltage characteristics of the driving transistor and current-voltage characteristics of the organic EL element. In the horizontal axis,

the direction of dropping with respect to the source potential of the driving transistor is the normal direction.

In the figure, current-voltage characteristics of the driving transistor and current-voltage characteristics of the organic EL element which correspond to two different gradation levels are shown, and the current-voltage characteristics of the driving transistor corresponding to a low gradation level is indicated by Vsig1 and the current-voltage characteristics of the driving transistor corresponding to a high gradation level is indicated by Vsig2.

In order to eliminate the impact of display defects due to changes in the source-to-drain voltage of the driving transistor, it is necessary to cause the driving transistor to operate in the saturation region. On the other hand, the pixel luminescence of the organic EL element is determined according to 15 the drive current. Therefore, in order to cause the organic EL element to produce luminescence precisely in accordance with the gradation level of video data, it is sufficient that the voltage remaining after the drive voltage (VEL) of the organic EL element corresponding to the drive current of the organic 20 EL element is deducted from the voltage between the source electrode of the driving transistor and the cathode electrode of the organic EL element is a voltage that can cause the driving transistor to operate in the saturation region. Furthermore, in order to reduce power consumption, it is preferable that the 25 drive voltage (VTFT) of the driving transistor be low.

Therefore, in FIG. **25**, the organic EL element produces luminescence precisely in accordance with the gradation level of the video data and power consumption can be reduced most with the VTFT+VEL that is obtained through the characteristics passing the point of intersection of the current-voltage characteristics of the driving transistor and the current-voltage characteristics of the organic EL element on the line indicating the boundary between the linear region and the saturation region of the driving transistor.

In this manner, the required voltage VTFT+VEL corresponding to the gradation levels for each color may be calculated using the graph shown in FIG. 25.

With this, power consumption can be further reduced.

It is to be noted that although the respective display devices according to Embodiments 2, 3, and 5 have a configuration in which a peak signal detecting circuit is provided, taken from the viewpoint of operating the driving transistor in the saturation region, it is possible to cause the organic EL element to produce luminescence precisely even without the peak signal 45 detecting circuit, by storing the voltage (VTFT+VEL) at the peak gradation level in a memory beforehand, as in Embodiments 1 and 4, and setting the voltage (VTFT+VEL) as the reference voltage at all times.

Furthermore, in Embodiment 1, the voltage margin setting 50 unit 175 outputs, to the high-side potential variable-voltage source 180, the reference voltage Vref1A to which the potential difference  $\Delta V$  detected by the high-side potential difference detecting circuit 170 has been added. In contrast, the display device 50 according to Embodiment 1 may be configured without a potential difference detecting circuit and with the potential of the detecting point M1 being inputted directly to the high-side potential variable-voltage source 180, as in the display device 200 according to Embodiment 3. Even with this configuration, it is possible to produce the same advantageous effects as with the display device 50 according to Embodiment 1.

Furthermore, in Embodiment 4, the high-side potential voltage margin setting unit 175A outputs, to the high-side potential variable-voltage source 180A, the reference voltage 65 VHref1 to which the potential difference  $\Delta$ VH detected by the high-side potential difference detecting circuit 170A has been

36

added, and the low-side potential voltage margin setting unit 175B outputs, to the low-side potential variable-voltage source 180B, the reference voltage VLref1 to which the potential difference  $\Delta$ VL detected by the low-side potential difference detecting circuit 170B has been added. In contrast, the display device 300 according to Embodiment 4 may be configured without a high-side potential difference detecting circuit and low-side potential difference detecting circuit and low-side potential difference detecting point  $M_A$  and detecting point  $M_B$  being inputted directly to the high-side potential variable-voltage source 180A and the low-side potential variable-voltage source 180A, as in the display device 200 according to Embodiment 3 Even with this configuration, it is possible to produce the same advantageous effects as with the display device 50 according to Embodiment 1.

Furthermore, in Embodiments 5, the voltage regulating unit may detect the potential difference between (i) the potential difference between the low-side potential of the monitor pixels measured by the voltage detection unit and the reference potential and (ii) the predetermined voltage, and regulate the power supplying unit in accordance with the detected potential difference.

Furthermore, in Embodiment 2, the signal processing circuit may change the first reference voltage Vref1 on a plural frame (for example, a 3-frame) basis instead of changing the first reference voltage Vref1 on a per frame basis.

With this, the power consumption occurring in the highside potential variable-voltage source 180 can be reduced because the potential of the first reference voltage Vref1 fluctuates.

Furthermore, the signal processing circuit may measure the potential differences outputted from the potential difference detecting circuit and the potential comparison circuit over plural frames, average the measured potential differences, and regulate the variable-voltage source in accordance with the average potential difference. Specifically, the process of detecting the potential at the detecting point (step S14) and the process of detecting the potential difference (step S15) in the flowchart shown in FIG. 12 may be executed over plural frames, and the potential differences for the plural frames detected in the process of detecting the potential difference (step S15) may be averaged in the process of determining the voltage margin (step S16), and the voltage margin may be determined in accordance with the average potential difference.

Furthermore, the signal processing circuit may determine the first reference voltage Vref1 and the second reference voltage Vref2 with consideration being given to an aged deterioration margin for the organic EL element 121. For example, assuming that the aged deterioration margin for the organic EL element 121 is Vad, the signal processing circuit 160 may determine the voltage of the first reference voltage Vref1 to be VTFT+VEL+Vdrop+Vad, and the signal processing circuit 260 may determine the voltage of the second reference voltage Vref2 to be VTFT+VEL+Vad.

Furthermore, although the switch transistor **124** and the driving transistor **125** are described as being P-type transistors in the above-described embodiments, they may be configured of N-type transistors.

Furthermore, although the switch transistor 124 and the driving transistor 125 are TFTs, they may be other field-effect transistors.

Furthermore, the processing units included in the display devices 50, 100, 200, 300, and 400 according to the above-described embodiments are typically implemented as an LSI which is an integrated circuit. It is to be noted that part of the processing units included in the display devices 50, 100, 200,

300, and 400 can also be integrated in the same substrate as the organic EL display units 110, 310, and 410. Furthermore, they may be implemented as a dedicated circuit or a general-purpose processor. Furthermore, a Field Programmable Gate Array (FPGA) which allows programming after LSI manufacturing or a reconfigurable processor which allows reconfiguration of the connections and settings of circuit cells inside the LSI may be used.

Furthermore, part of the functions of the data line driving circuit, the write scan driving circuit, the control circuit, the peak signal detecting circuit, the signal processing circuit, and the potential difference detecting circuit included in the display devices 50, 100, 200, 300, and 400 according to the exemplary embodiments may be implemented by having a processor such as a CPU execute a program. Furthermore, the 15 exemplary embodiments may also be implemented as a display device driving method including the characteristic steps implemented through the respective processing units included in the display devices 50, 100, 200, 300, and 400.

Furthermore, although the foregoing descriptions exemplify the case where the display devices **50**, **100**, **200**, **300**, and **400** are active-matrix organic EL display devices, one or more exemplary embodiments may be applied to organic EL display devices other than that of the active-matrix type, and may be applied to a display device other than an organic EL display device using a current-driven luminescence element, such as a liquid crystal display device.

Furthermore, for example, a display device according to the present disclosure is built into a thin flat-screen TV such as that shown in FIG. **26**. A thin, flat TV capable of high-accuracy image display reflecting a video signal is implemented by having the display device according to the present disclosure built into the TV.

Each of the structural elements in each of the above-described embodiments may be configured in the form of an 35 exclusive hardware product, or may be realized by executing a software program suitable for the structural element. Each of the structural elements may be realized by means of a program executing unit, such as a CPU and a processor, reading and executing the software program recorded on a 40 recording medium such as a hard disk or a semiconductor memory.

The herein disclosed subject matter is to be considered descriptive and illustrative only, and the appended Claims are of a scope intended to cover and encompass not only the 45 particular embodiment(s) disclosed, but also equivalent structures, methods, and/or uses.

## INDUSTRIAL APPLICABILITY

One or more exemplary embodiments described herein are particularly useful as an active-type organic EL flat panel display.

The invention claimed is:

- 1. A display device, comprising:
- a power supply configured to output at least one of a highside output potential and a low-side output potential;
- a display in which pixels are arranged and which receives power supply from the power supply;
- a voltage detector configured to detect at least one of a high-side applied potential to be applied to at least one of the pixels in the display and a low-side applied potential to be applied to at least one of the pixels in the display which is the same as or different from the at least one of 65 the pixels to which the high-side applied potential is to be applied;

38

a voltage regulator configured to regulate at least one of the high-side output potential and the low-side output potential outputted by the power supply to set, to a predetermined potential difference, a potential difference between (i) at least one of the high-side applied potential and the low-side applied potential and (ii) a reference potential; and

at least one of:

- a high-side potential detecting line having one end connected to the at least one pixel for which the high-side applied potential is detected and the other end connected to the voltage detector, for transmitting the high-side applied potential to the voltage detector; and
- a low-side potential detecting line having one end connected to the at least one pixel for which the low-side applied potential is detected and the other end connected to the voltage detector, for transmitting the low-side applied potential to the voltage detector.
- 2. The display device according to claim 1,
- wherein the at least one pixel for which the high-side applied potential is detected and the at least one pixel for which the low-side applied potential is detected are different pixels.
- 3. The display device according to claim 1,
- wherein at least one of (i) the at least one pixel for which the high-side applied potential is detected and (ii) the at least one pixel for which the low-side applied potential is detected is plural in number.
- **4**. The display device according to claim **3**, wherein the voltage regulator is configured to:
- select at least one applied potential from:
  - a lowest applied potential among high-side applied potentials detected by the voltage detector; and
  - a highest applied potential among low-side applied potentials detected by the voltage detector; and

regulate the power supply based on the selected at least one applied potential.

- 5. The display device according to claim 1,
- wherein the voltage detector is further configured to detect at least one of the high-side output potential and the low-side output potential which are outputted by the power supply, and
- the voltage regulator is configured to regulate at least one of the high-side output potential and the low-side output potential which are outputted by the power supply, in accordance with a first potential difference which is at least one of (i) a potential difference between the highside output potential and the high-side applied potential and (ii) a potential difference between the low-side output potential and the low-side applied potential.
- 6. The display device according to claim 5,
- wherein the voltage regulator is configured to perform the regulation that realizes an increasing function relationship between (i) at least one of a potential difference between the high-side output potential and the reference potential and a potential difference between the low-side output potential and the reference potential and (ii) the first potential difference.
- 7. The display device according to claim 1,

55

60

- wherein the voltage detector is further configured to detect at least one of (i) a potential in a high-side potential current path connecting the power supply and a high potential side of the pixels and (ii) a potential in a lowside potential current path connecting the power supply and the low potential side of the pixels; and
- the voltage regulator is configured to regulate at least one of the high-side output potential and the low-side output

40
10. The display device according to claim 9,

potential that are outputted by the power supply, in accordance with a second potential difference which is at least one of (i) a potential difference between the potential in the high-side potential current path and the high-side applied potential and (ii) a potential difference between the potential in the low-side potential current path and the low-side applied potential.

8. The display device according to claim 7,

wherein the voltage regulator is configured to perform the regulation that realizes an increasing function relationship between (i) at least one of a potential difference between the high-side output potential and the reference potential and a potential difference between the low-side output potential and the reference potential and (ii) the second potential difference.

9. The display device according to claim 1,

wherein each of the pixels includes:

a driving element having a source electrode and a drain electrode; and

a luminescence element having a first electrode and a second electrode, the first electrode being connected to one 20 of the source electrode and the drain electrode of the driving element,

the high-side applied potential is applied to one of the second electrode and the other of the source electrode and the drain electrode, and

the low-side applied potential is applied to the other of the second electrode and the other of the source electrode and the drain electrode. 10. The display device according to claim 9, wherein the pixels are arranged in rows and columns,

the display further comprises a first power source line and a second power source line, the first power source line connecting the others of the source electrode and the drain electrode of the respective driving elements of adjacent pixels in at least one of the row direction and the column direction, and the second power source line connecting the second electrodes of the respective luminescence elements of adjacent pixels in the row direction and the column direction, and

the pixels receive power from the power supply via the first power source line and the second power source line.

11. The display device according to claim 10,

wherein the second electrode and the second power source line are part of a common electrode that is common to the pixels, and are electrically connected to the power supply so that a potential is applied to the common electrode from a periphery of the common electrode.

12. The display device according to claim 11,

wherein the second electrode comprises a transparent conductive material including a metal oxide.

13. The display device according to claim 9,

wherein the luminescence element is an organic electroluminescence (EL) element.

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